# TSTE12 Design of Digital Systems Lecture 6 

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## Agenda

- Practical issues
- Handins
- Audio codec function and interface
- High abstraction level modelling


## TSTE12 Deadlines Y,D,ED

- Final version of Requirement specification today 8 September
- Initial version design sketch and project plan tuesday 12 September
- Design sketch (proper document) describe block structure and functions
- Project plan describe activities (what, who and when)
- Weekly meetings should start
- Internal weekly meeting with transcript sent to supervisor
- Lab 2 soft deadline Wednesday 13 September at 21.00
- Lab 2 results will be checked after project end


## TSTE12 Deadlines MELE, erasmus

- First project meeting no later than Monday 11 September
- Tuesday 12 September: First version of requirement specification
- Wednesday 13 September 21.00: Lab 1 deadline
- Pass required to be allowed continued project participation


## Handin (homework), Individual!

- $1^{\text {st }}$ handin deadline Monday 18 September
- Available on web from Monday 11 September
- Submit answers using Lisam assignment function (individual work!!!)
- Theory question answers entered direct as text (see assignments on web)
- Use your own home directory for code answer testing (Not lab group directory)
- Use ~/TSTE12/
- Use a special terminal window when working with handins
module load TSTE12 ; TSTE12handin


## Individual handin task, cont.

- Create handin code answers using a plain text editor
- emacs, vim, or the built-in editor in modelsim
- See in the tutorial how start and use modelsim
- Upload the answers of the coding tasks onto Lisam (TSTE12 course room submission)
- Remember to compile and simulate the design
- Will use source code for checking handin results
- Do not use handin directory for anything else but handin code and answers you make yourself!


## Individual handin, cont.

- Hand-ins are individual work!
- Ask me if there are questions about the handin
- Hand-ins are checked automatically (using scripts)
- Make sure all names and types are correct in code answers
- Datatype bit is not the same as std_logic!
- Test your code! Do not assume that you have written correct code.
- NOTE: Do NOT use hdl-designer (do not start the software using TSTE12lab or TSTE12proj)
- See modelsim tutorial on exercise page
- www.isy.liu.se/edu/kurs/TSTE12/kursmaterial


## DE2-115 board components

- Audio codec used to input/output analog audio signal
- Codec function
- Codec can be used in multiple configurations
- Contains clock generators, A/D, D/A and filters
- Loopback, volume control
- Codec configuration
- Default configuration defined in documentation


## Individual

- DE2-115 FPGA default design
- DE2-115 loads a default design at power on
- Microprocessor design running (NIOS II Soft process, i.e. written i VHDL)
- Check switches SW3 downto SW0 to select what to do with the SRAM contents and Codec init
- Infinite loop: read switches, update LEDs, updates 7-segment display.
- Help text shown on VGA screen
- Default design is not the standard design described in the DE2-115 user manual
- Do not depend on power-on defaults
- Allows configuration of memory and codec for testing


## Codec programming (I2C)

- Codec configured using an I2C bus
- General structure
- Multichip bus (all chips connect to the same pins)

- Pullup
- Only assign 'Z' or '0' on pins
- Pullup will translate ' $Z$ ' to ' 1 ' if other chips does not sink that pin
- Separate output and input signals to/from pin

- Values do not change immediately (slow voltage changes)


## I2C protocol

- Bidirectional protocol
- Wired-and using pull-up
- Send byte by byte
- MSB first

- FPGA work as master
- Slave (codec) responds with ack after each byte
- Pulls down SDA in ack cycle
- Simple solution: assume ack (do not check)


## Numeric calculations

- Bit-vectors (and std_logic_vectors) does not directly correspond to a value
- "1011" could mean 11 in decimal (unsigned), or -5 in decimal (2's complement)
- Datatypes are included in supporting packages to enable arithmetic on bit-vectors
- ieee.numeric_bit.all
- ieee.numeric_std.all
- Must use defined types signed or unsigned to allow calculations
- Same definitions as bit_vector and std_logic_vector
- Can copy values between types due to same element type


## Numeric calculations example

- Counter incrementing 3-bit count value each clock cycle
- Asynchronous reset
library ieee; use ieee.numeric_bit.all;
entity INL3_KB is port ( C : in bit; R : in bit; Q : out bit_vector(1 to 3)); end entity;
architecture KB of INL3_KB is begin

```
process(C,R)
```

        variable count : unsigned(1 to 3 );
        begin
        if \(R=\) ' 1 ' then
            count := (others => '0');
        elsif C'event and ( \(C=\) '1') then
            count := count + 1;
            end if;
            \(\mathrm{Q}<=\) bit_vector(count);
            end process;
                            end architecture;
    
## Including integers

- Integers can be used for synthesis
- If synthesis tool cannot figure out the limits, the result is 32-bit arithmetic
- Subtypes (limiting range) help to reduce hardware and catch unexpected use
- Integers will be implemented as bitvectors
- Either unsigned or signed (2’s complement)
- Translation between integer and bitvectors exist x_signed := to_signed(y_int,x_signed’size);
- Translation other way around (unsigned to integer value) y_int = to_integer(x_signed);


## Another aspect of signal assignment

- One signal can be assigned from different parts of the code
- Support multiple entities driving the same wire
- Example: Databus in a computer connecting multiple memories and CPU
- Modelling must be strict and clear
- Same result independant of simulator tool
- Should not be able to detect the order the processes where calculated
- Not all data types support multiple sources for the value


## Multiple assignment on one signal

- Each process containing a signal assignment will have a driver in the simulator generating a contribution to the final signal value
- Concurrent signal assignments will have one driver each
- Processes only have one driver for each signal (even with multiple assignment)
- The signal update seen before is done individually on each driver
- One driver does not know anything about other drivers
- When the value of a signal is fetched, the contributions from the different drivers current values are collected.
- The resulting signal value depends on the definition of how to combine the values from the different drivers, using a resolution function


## Example of data types supporting multiple drivers

- Signals driven by multiple drivers must be resolved
- Use a special function that resolves multiple drivers
- Resolution function
- Example: Wired-OR
- signal X1 : WIRED_OR Bit;
- subtype STD_LOGIC is RESOLVED STD_ULOGIC;
- signal Y2 : STD_LOGIC;
- RESOLVED is the resolution function name
- Called every time the value of the signal is calculated
- Gets all driver values as input

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## Example of implementing Multivalued logic in VHDL

- Alternative to data type BIT but simpler than std_logic

Type MVL4 is ('X', '0', '1', 'Z');
Type MVL4_VECTOR is array(NATURAL range <>) of MVL4;

- X leftmost to make it the initial value unless explicitly initialized in the code


## Multidriver signals

- Requires a resolution signal
- Different combinations possible
- X always overrides others
- 0 and 1 at the same time gives $X$
- $Z$ and $Z$ gives $Z$


## Resolution function definition

Subtype DotX is wiredX MVL4;

- WiredX is the name of the resolution function

Function WiredX (V:MVL4_VECTOR) return MVL4;

- Where V is a vector containing all values of all drivers of a signal


## Resolution function implementation

- Implement as a loop and lookup table

```
Function wiredX (V: MVL4_VECTOR) return MVL4 is
    Variable result: MVL4:= 'Z';
Begin
    For i in V'RANGE loop -- range not known in advance
        Result = table_WIREDX(result,V(i));
        Exit when result = 'X';
        End loop;
        Return result;
    End wiredX;
```


## Resolution function impl., cont.

- Check of X in loop is not necessary, but speed up simulation
- Table should then look like:

```
Type MVL4_TABLE is array (MVL4, MVL4) of MVL4;
Constant table_WIREDX : MVL4_TABLE :=
--
-- X 0 1 Z
--
    (('X', 'X', 'X', 'X'), -- X
    ('X', '0', 'X', '0'), -- 0
    ('X', 'X', '1', '1'), -- 1
    ('X', '0', '1', 'Z')); -- Z
```


## Resolution function impl. Cont.

- Table lookup may be used for most functions
- Not possible to know the order of the value in $V$, may therefore require a more complex algorithm


## Bus data type

Type busX is array (Natural range $<>$ ) of DotX;

- However, a new data type requires all logic operations to be specified
- Complicated
- Better approach: conversion function
- Only read bus using a call to a Sense function

Function Sense (value : busX) return bit_vector;

- Only assign value to the bus using the Drive function

Function Drive (value : bit_vector) return busX

## Algorithmic level development

- Specification in many cases in natural language
- Ambigous description in many cases
- Want an executable specification
- Allows testing of the behavior the description describes
- Use VHDL to capture the specification
- Use the full language capabilities
- Description not intended for synthesis


## Process Model Graph (PMG)

- Typical example
- Arcs describes signals with names and delays

- example process 4 to 1
- VHDL Code example: S <= xxx after DEL_S;
- Physical or functional partitioning
- Single process may map to multiple hardware units
- Multiple process may map to single hardware unit


## Graph Elements

- Two types of signals
- Triggering signal, put in sensitivity list
$\longrightarrow$ Sampled signal

- Signals without delay information has a delay of one delta-t
- Signals may be driven by multiple processes. Requires a resolution function
- Signals may be bidirectional. Requires also a resolution function


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## Example approach

- Map groups of sentences onto VHDL processes
- Assign each process an activity list
- Develop VHDL code that implements each activity


## Example: Serial to Paralell converter

- English text description
- The 8-bit parallel word (PARIN) is loaded into the converter when the control signal LD makes a zero to one transition At this time the status signal BUSY is set high. The data is shifted out serially at a rate controlled by the input shift clock SHCLK. Shifting occurs at the rise of the clock. BUSY remains high until shifting is complete. While BUSY is high, no further loads will be accepted.
- Note some sentences are shared between functions
- Two processes: LOAD and SHIFT


## Serial to Paralell converter, cont.

- LOAD: (a) 8-bit parallel word (PARIN) load when LD makes a zero to one transition. Set BUSY high. (b) BUSY remains high until shift complete. No new loads while BUSY high
- SHIFT: (a) Data shifted out controlled by rising edge of SHCLK.
(b) BUSY remain high until shift complete


```
entity PAR_TO_SER is
    port(LD,SHCLK: in BIT;
        PARIN: in BIT_VECTOR(0 to 7);
        BUSY: inout BITT := '0';
        SO: out BIT);
end PAR_TO_SER;
```


## PMG version

- Corresponding code based on processes
- PMG defines interface of each process + signals between the processes
- Code start by defining processes and comments about activities

```
architecture TWO_PROC of PAR_TO_SER is
    signal SH_COMP: BIT :='0';
    signal PREG: BIT_VECTOR(0 to 7);
begin
    LOAD: process(LD,SH_COMP)
    begin
        ---- Activities:
            ----1)Register Load
            ----2)Busy Set
            ----3)Busy Reset
    end process LOAD;
    SHIFT: process(BUSY,SHCLK)
            variable COUNT: INTEGER;
            variable OREG: BIT_VECTOR(0 to 7);
    begin
            ----Activities:
            ----1)Shift Initialize
            ----2)Shift
            ----3)Shift Complete
    end process SHIFT;
end TWO_PROC;
```


# PMG -> Code 

SHIFT: process(BUSY, SHCLK)
variable COUNT: INTEGER;
variable OREG: BIT_VECTOR(0 to 7);
begin
----Activities:
if BUSY'EVENT and

- Each process has a check for an event, and then a part that execute the data operations


## BUSY $=$ '1' then

----1)Shift Initialize
COUNT : = 7;
OREG := PREG;
SH_COMP <= '0';
end if;
if SHCLK'EVENT and SHCLK= '1'and BUSY='1' then ----2)Shift SO<=OREG(COUNT); COUNT := COUNT - 1; ----3)Shift Complete if COUNT < 0 then

SH_COMP <= '1'; end if;
end if;
end process SHIFT;

```
LOAD:process(LD,SH_COMP)
begin
    ---- Activities:
    if LD'EVENT and LD='1'
        and BUSY='0' then
        ----1)Register Load
        PREG <= PARIN;
        ----2)Busy Set
        BUSY <= '1';
    end if;
        if SH_COMP'EVENT
                and SH_COMP='1' then
                ----3)Busy Reset
                BUSY <= '0';
    end if;
end process LOAD;
```

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## Timing example

- New model: Buffered register
- Loaded on rise of the strobe (STRB)
- English description:
- The register is loaded on the rise of the strobe (STRB), and assuming that the output buffers are enabled, the output of the buffers will change $t_{S D}$ nanoseconds later. The enable condition for the register buffer is the AND of the DS1 and invers of DS2 inputs. Any change in the enable condition will cause the outputs to change $\mathrm{t}_{\mathrm{ED}}$ nanoseconds later.


## Timing example, cont.

- Three processes: PREG, ENABLE, OUTPUT
- Add delay on wires

$$
\begin{aligned}
& \mathrm{t}_{\mathrm{SD}}=\mathrm{STRB} \_ \text {DEL }+ \text { ODEL } \\
& \mathrm{t}_{\mathrm{ED}}=\mathrm{EN} \mathrm{\_DEL}+\mathrm{ODEL}
\end{aligned}
$$



## Timing example, cont.

PREG: process(STRB) begin if (STRB = '1') then
entity BUFF_REG is generic(

STRB_DEL,EN_DEL,ODEL: TIME); port

DI: in BIT_VECTOR(1 to 8); STRB: in BIT;
DS1: in BIT;
NDS2: in BIT; DO: out BIT_VECTOR(1 to 8)); REG <=DI after STRB_DEL; end if;
end process PREG;
end BUFF_REG;
ENABLE: process(DS1,NDS2) begin ENBLD <= DS1 and not NDS2 after EN_DEL;
end process ENABLE;
OUTPUT: process(REG,ENBLD)
begin
if (ENBLD = '1') then DO <= REG after ODEL; else

DO <= "11111111" after ODEL;
end if;
end process OUTPUT;
end THREE_PROC;

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## Process complexity trade-off

- Number of signals
- Many signals => slow simulation
- Large processes
- Complex behavior may not match specification
- Ease of mapping to hardware
- More processes may simplify mapping


## Checking timing

- Additional requirements
- DI stable SUT ns before STRB rise
- DI stable HT ns after STRB rise
- STRB minimum high duration MPW ns
- Implement checks using assert statements

```
assert not (not STRB'stable and (STRB = '1')
    and not DI'stable(SUT))
    report "Setup Time Failure";
```

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## Timing Check placement

- Tests in architecture must be copied between architectures
- May introduce errors
- If changed, many architectures must be changed
- Solution: Place checks in the entity
- Check always executed, independent of selected architecture


## Timing check example

```
Entity BUFF_REG is
    Generic (STRB_DEL, EN_DEL, ODEL,SUT,HT,MPW: TIME);
    Port (DI: in bit_vector(1 to 8);
            STRB : in bit ; DS1 : in bit;
            NDS2 : in bit;
            DO : out bit_vector(1 to 8));
    Begin
        Assert STRB'stable or (STRB = '0') or DI'stable(SUT)
            Report "Setup time Failure";
    Assert STRB'delayed(HUT)'stable or
            (STRB'delayed(HT) = '0') or DI'STABLE(HT)
            Report "Hold Time Failure";
        Assert STRB'stable or (STRB = '1') or
            STRB'delayed'stable(MPW)
            Report "Minimum pulse width failure";
```

End BUFF_REG;

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