

# Test benches

- Lab 3
  - DMA
  - The task
  - (Bursts)
- Test benches

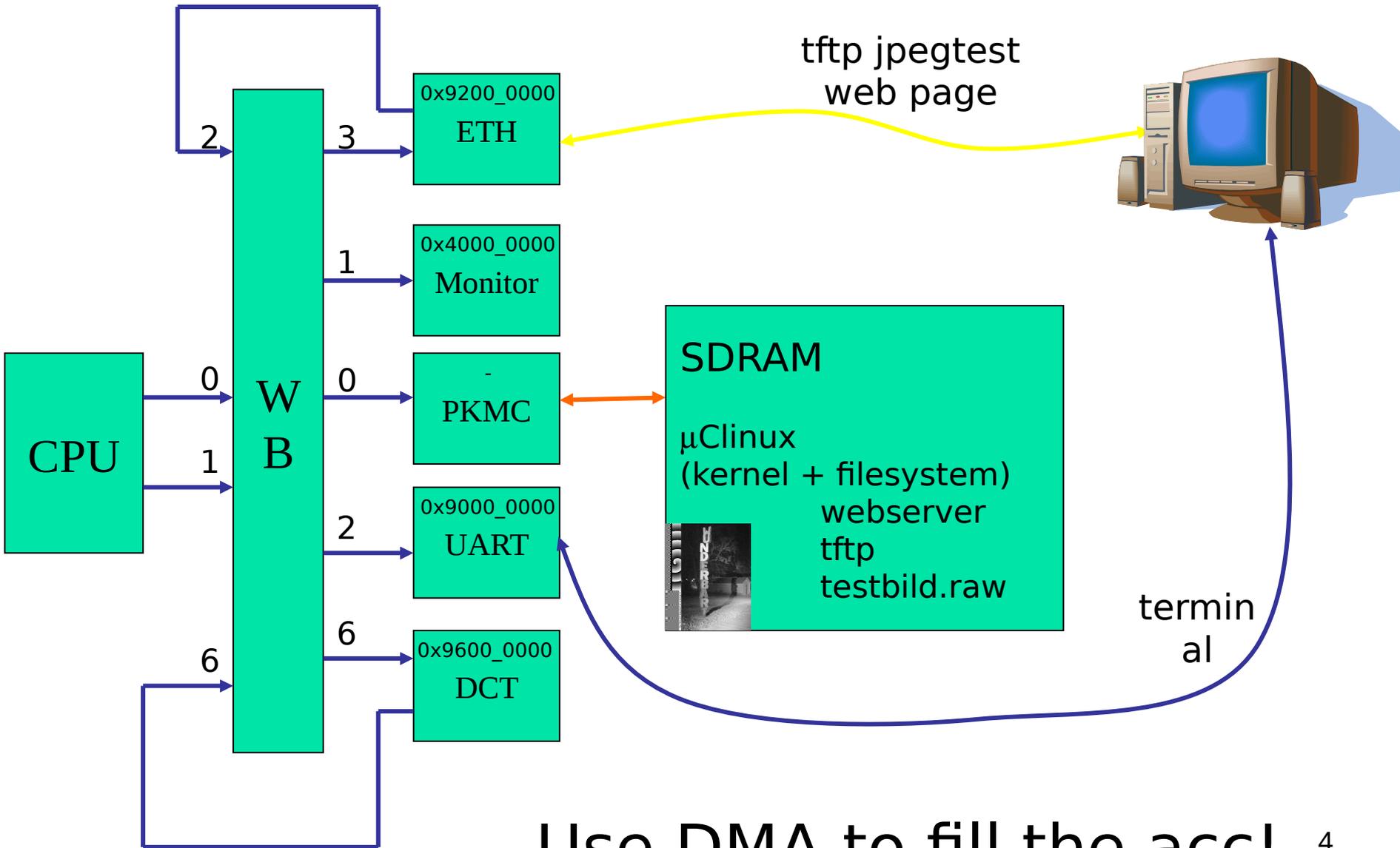
# Schematic diagrams in the lab reports

- Alternatives:
  - Open Office diagram tool (bletch)
  - Inkscape (potentially very nice looking, very cumbersome though)
  - Dia (decent, if you have RTL library for it)
  - TikZ (if you really like latex)
  - MS Paint (well, not really)
  - Hand drawn schematic from whiteboard/paper
  - Visio (if you have a license for it)

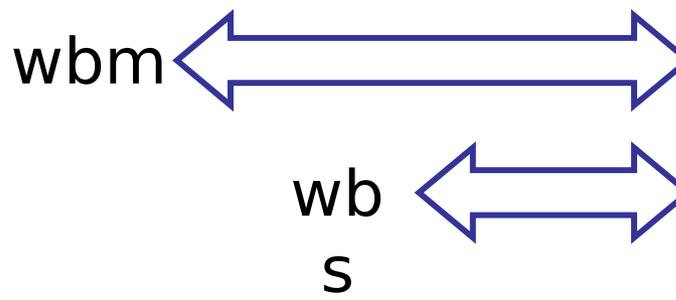
# schematic\_gui

- [http://github.com/ehliar/schematic\\_gui](http://github.com/ehliar/schematic_gui)
- `/site/edu/da/TSEA44/bin/schematic_gui`
  - (Can probably be started as `schematic_gui` after `module add TSEA44`)
  - Custom tool for digital circuit diagrams created by the examiner.
    - It works for me™
  - See tutorial at [https://github.com/ehliar/schematic\\_gui/blob/master/tutorial/tutorial.md](https://github.com/ehliar/schematic_gui/blob/master/tutorial/tutorial.md)

# Lab 3 - DMA

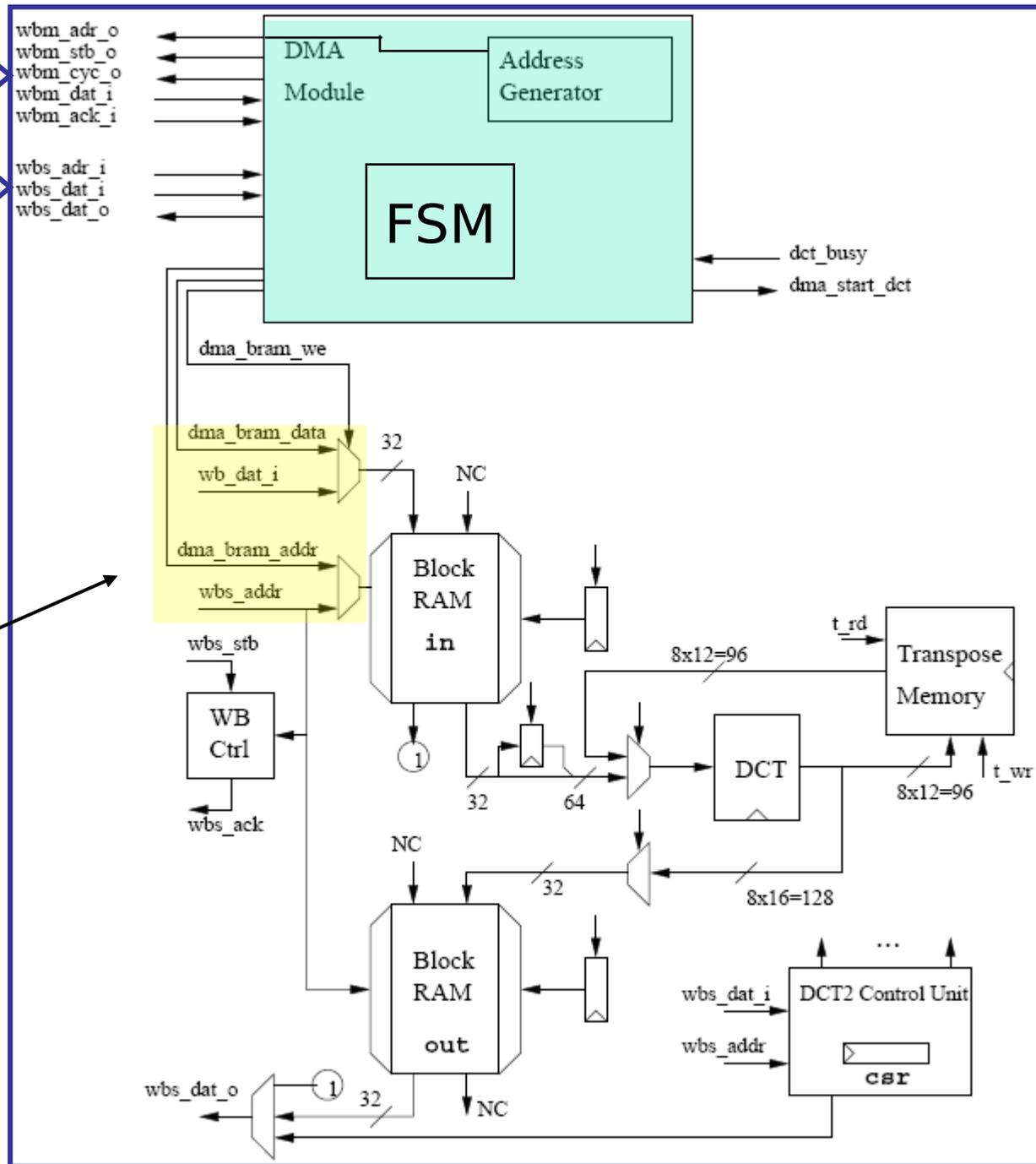


Use DMA to fill the acc!



# Proposed architecture

- Design FSM
- Change here
- Modify jpegfiles



# Address generation



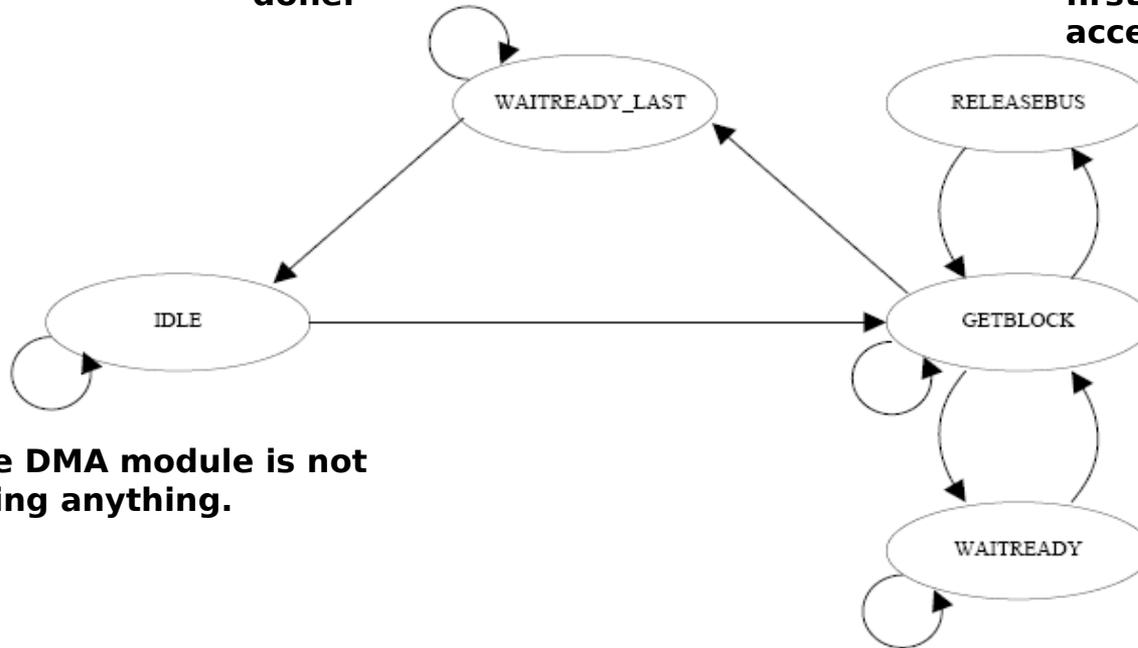
testbild.raw

- We want to transfer block by block (8x8)
- Address generator must know format (width,height) of image

# State diagram

Same as WAITREADY except that we go to the IDLE state when done.

The DMA accelerator has to release the bus regularly so that other components can access it. Do it for every line you read. When we finish the first block, we start the DCT accelerator.



The DMA module is not doing anything.

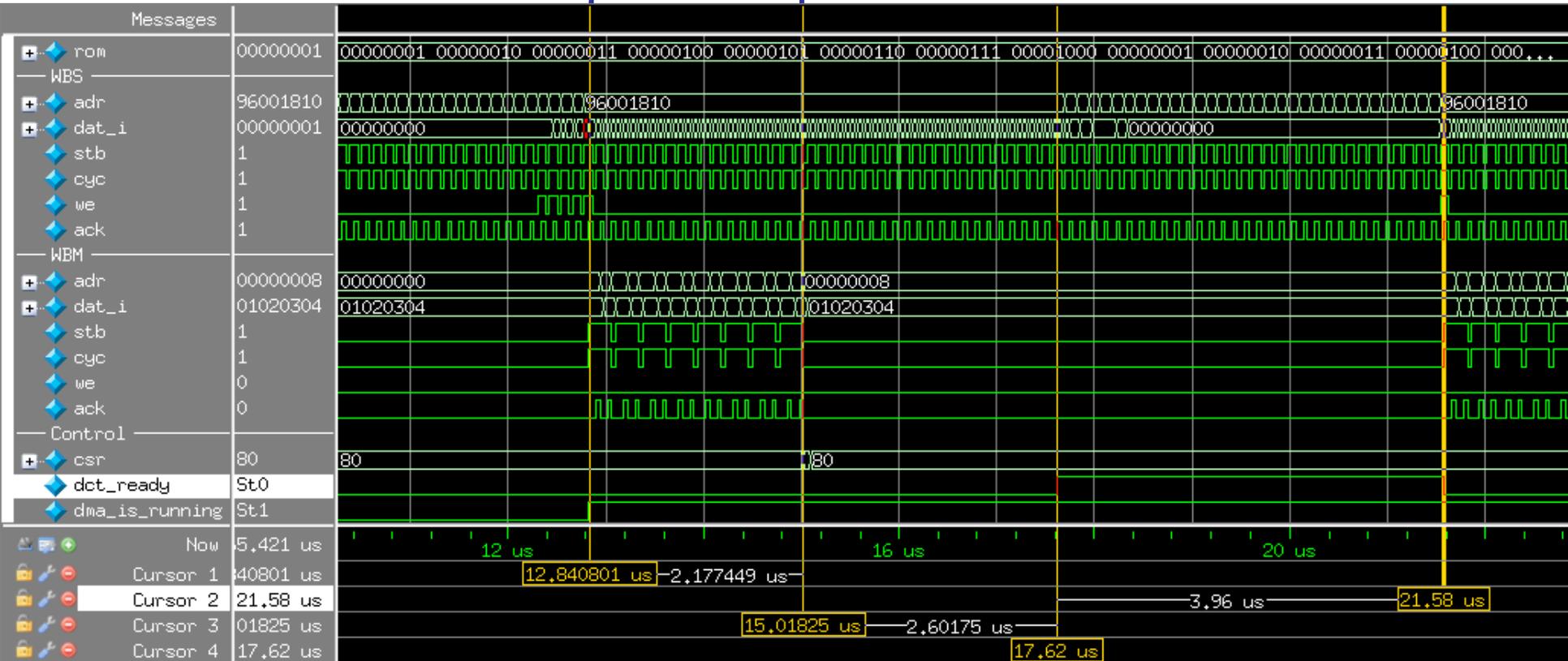
The DMA module is fetching an 8x8 block. Once the block is fetched we go to the WAITREADY state and start the DCT transform.

In this state we wait until the program tells us that it has read the result of the transform by writing to the control register

# A measurement make sim\_jpeg

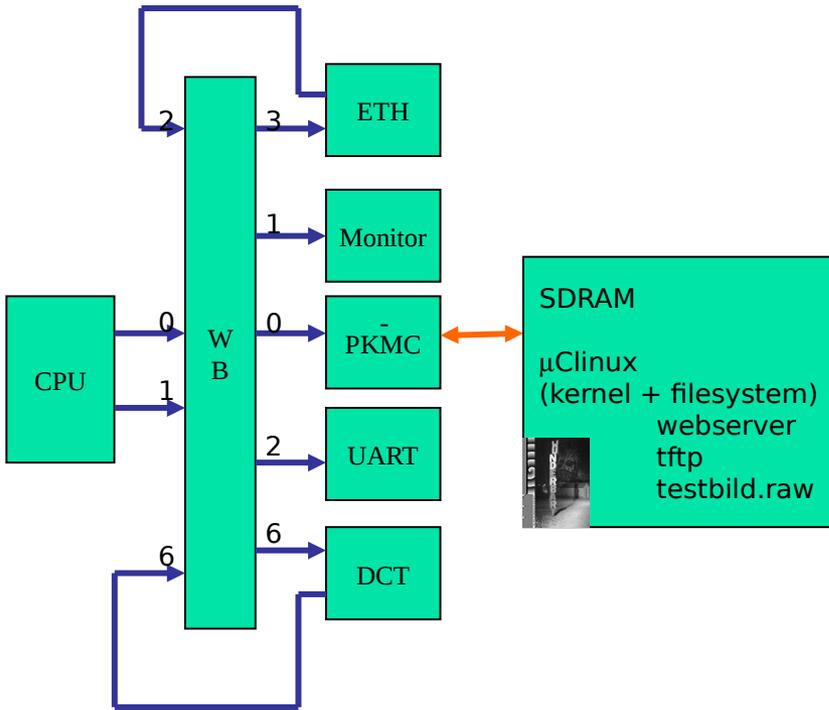
Copy 16 words  
from **SDRAM**  
to **DCT (DMA)**

Copy 32 words  
from **DCT**  
to **SDRAM**

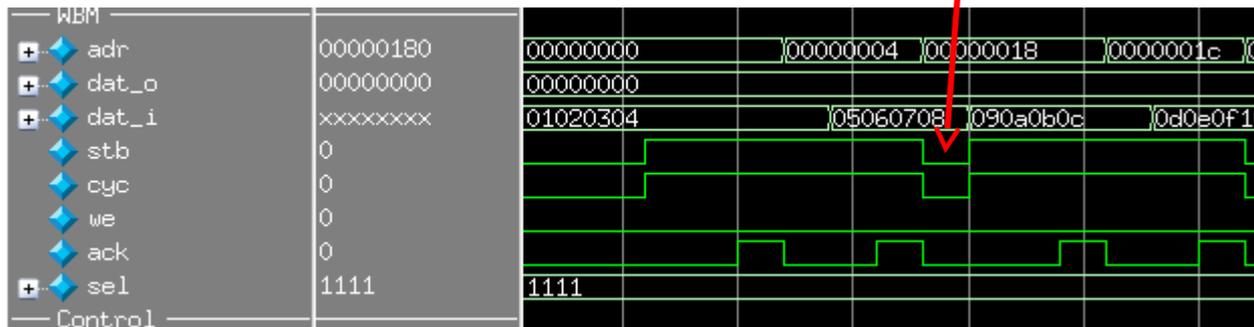


DCT

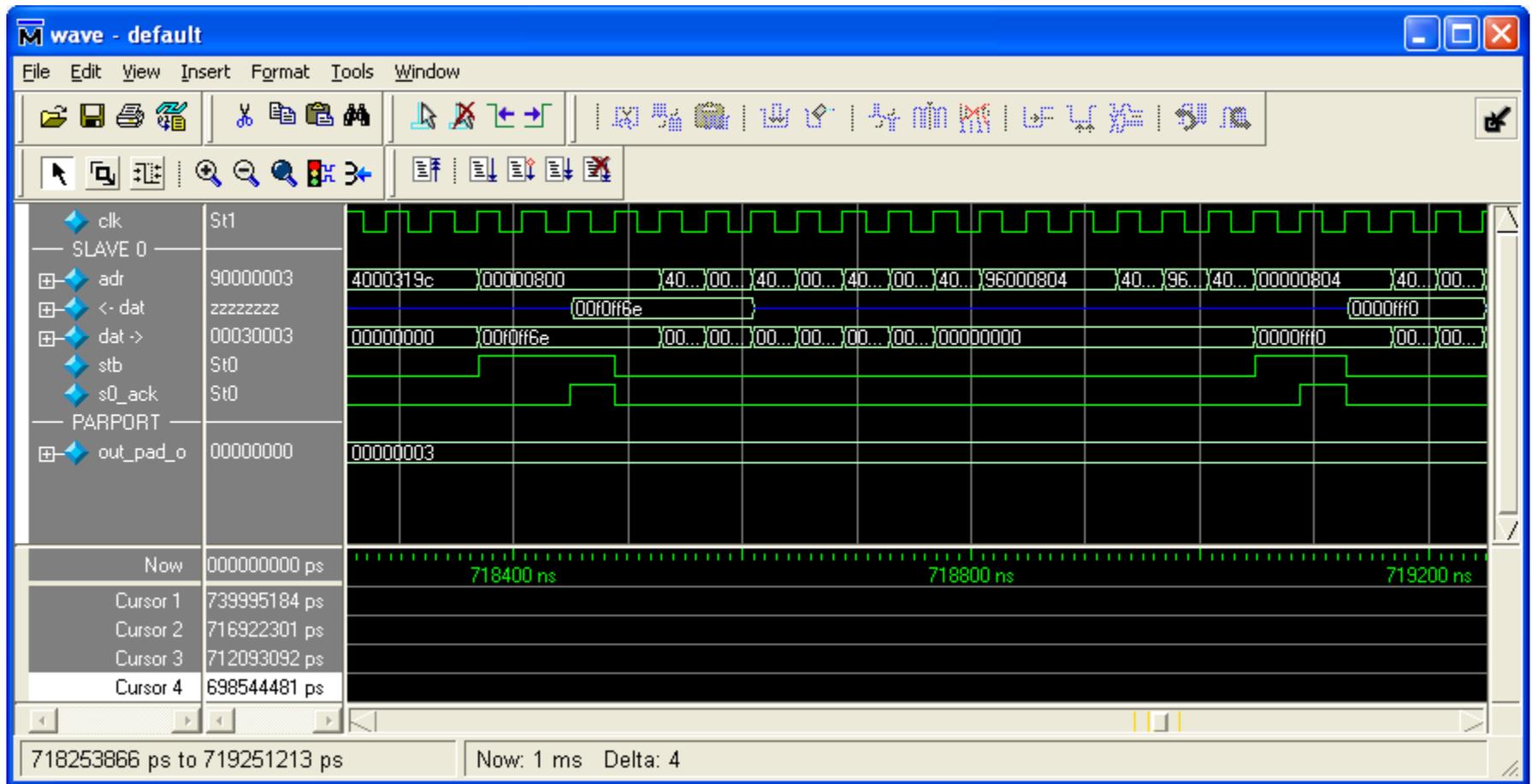
# A closer look at the DMA



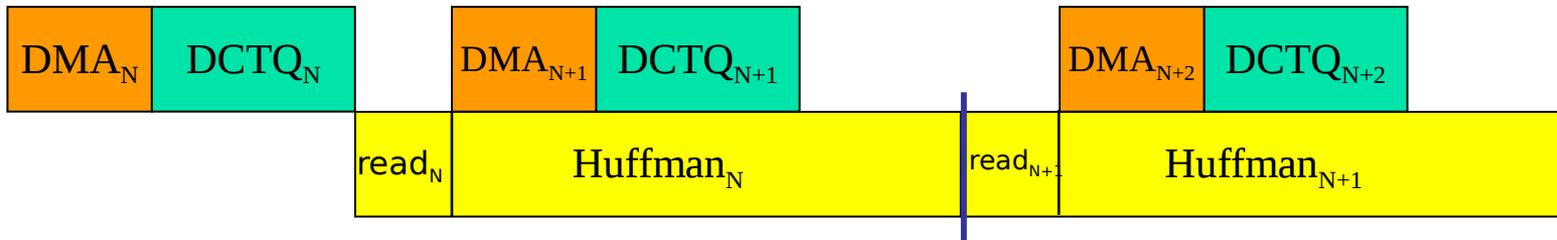
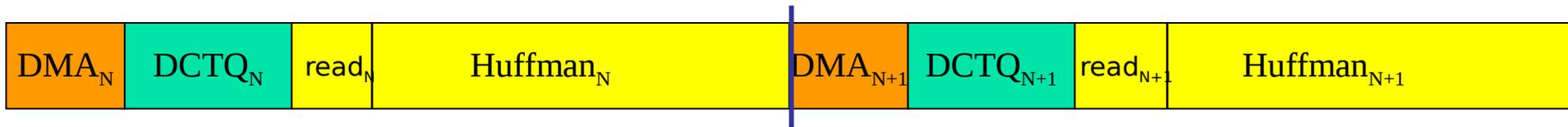
Release bus for  
m0, m1, m2



# DCT => Mem (Software)

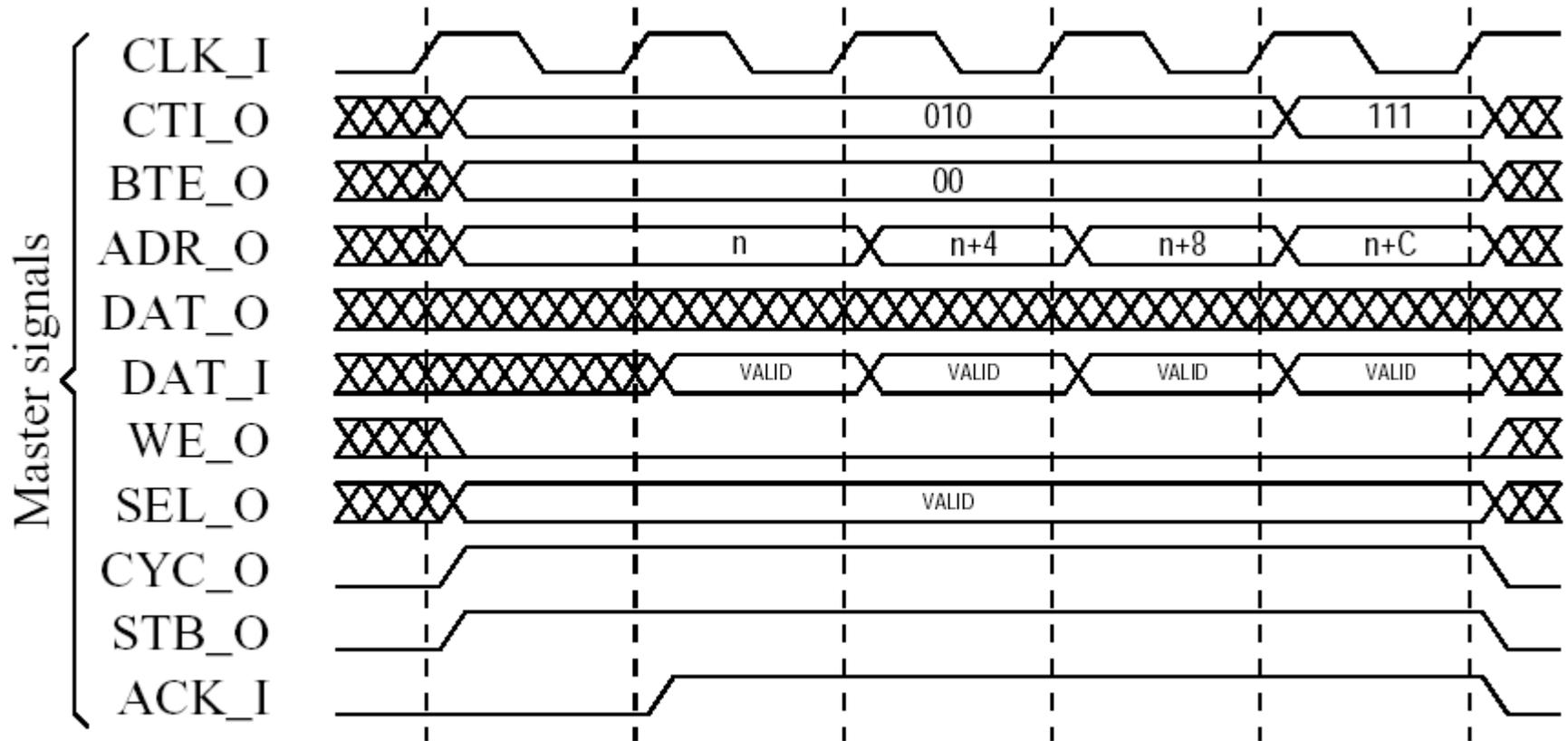


# A hint



How long time do these blocks take?

# Burst Read



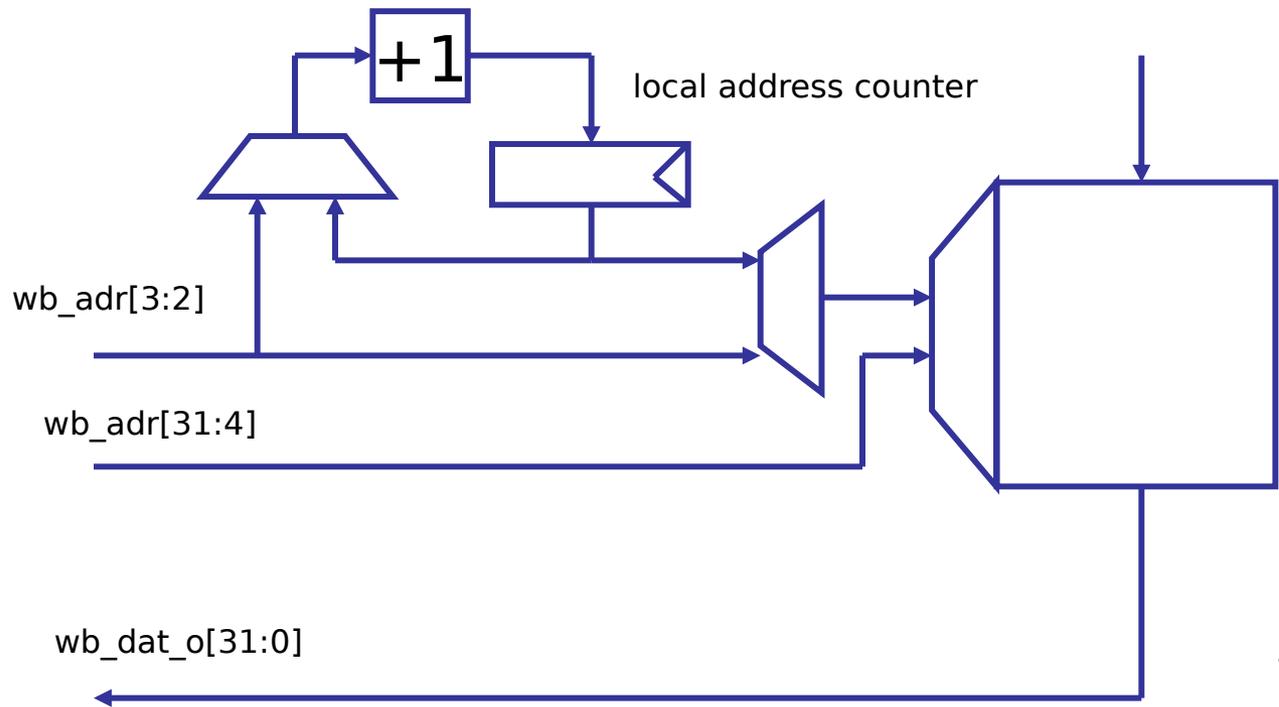
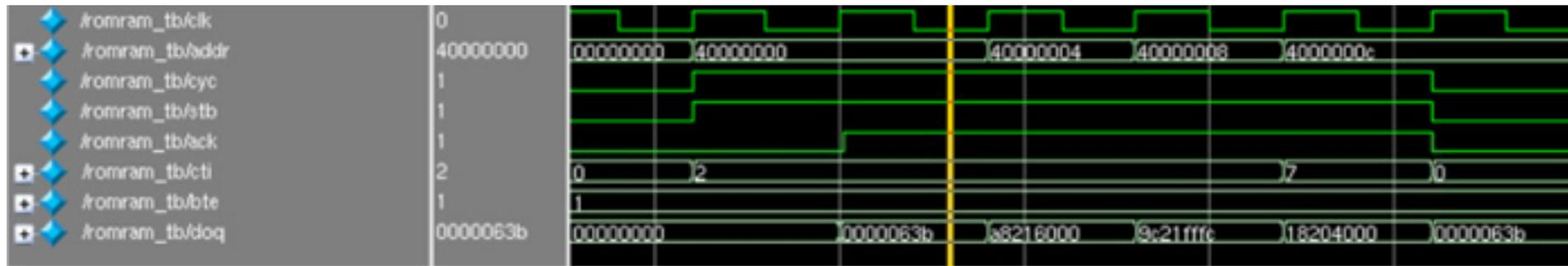
# burst - cycle types

Signal group	Value	Description
cti	000	Classic cycle
	001	Constant address burst cycle
	010	Incrementing burst cycle
	011-110	<i>Reserved</i>
	111	End of burst
bte	00	Linear burst
	01	4-beat wrap burst
	10	8-beat wrap burst
	11	16-beat wrap burst

# Burst access

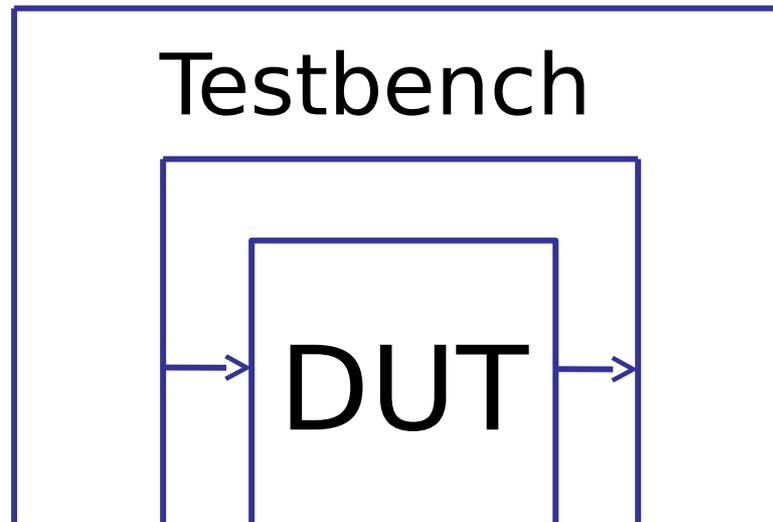
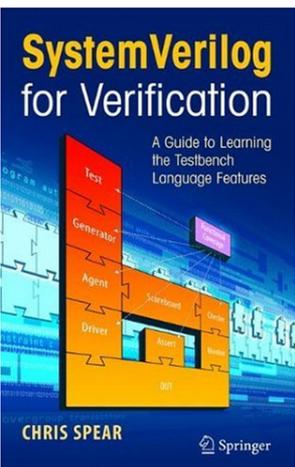
- Note: Only the SRAM memory controller in the Leela memory controller has burst support.

# Changes in the slave

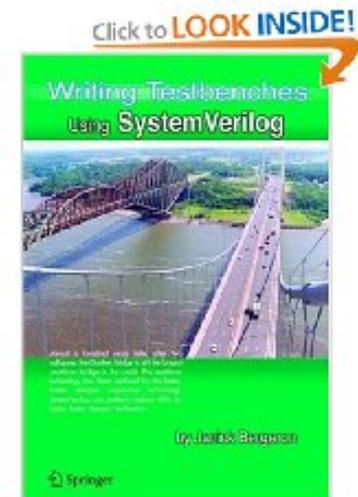


# Testbenches

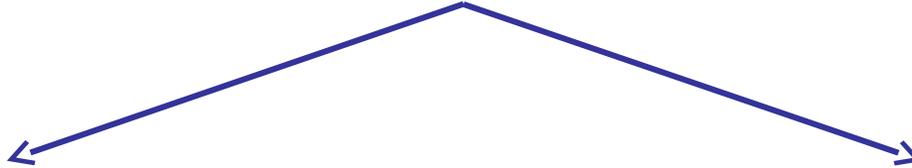
Spear, Chris:  
*System Verilog  
for verification.*  
Springer



Bergeron, Janick:  
*Writing testbenches  
using System Verilog.*  
Springer



# Testbenches



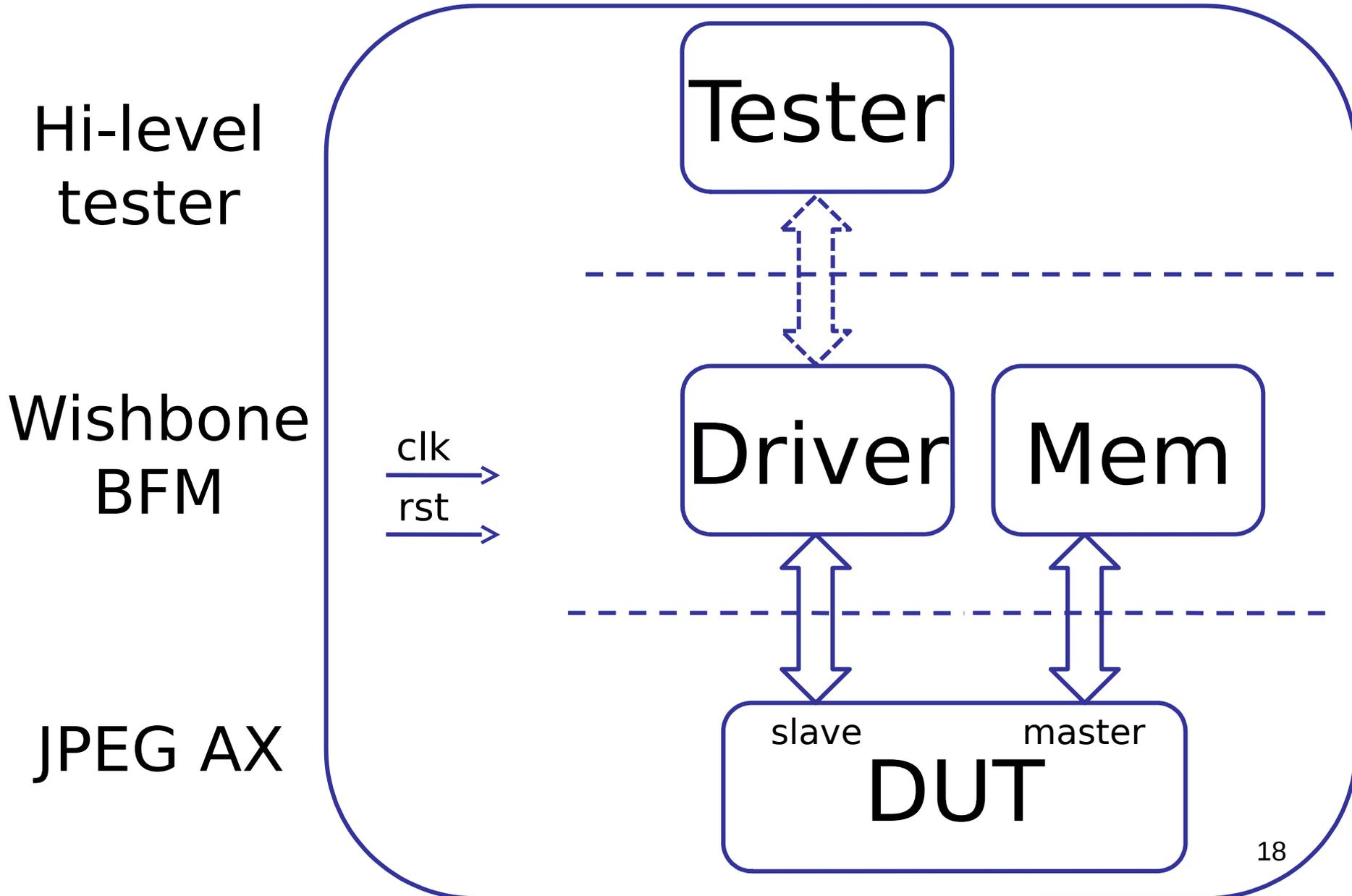
Like an FSM  
(same as DUT)

- complicated to design
- hard to test timing
- hard to test flow

Like High-Level Software  
(very different from DUT)

- easy to design
- easy to test timing
- easy to test flow

# An example : a TB for your design



# Testbench : top level

```
module jpeg_top_tb();
    logic        clk = 1'b0;
    logic        rst = 1'b1;
    wishbone wb(clk,rst), wbm(clk,rst);

    initial begin
        #75 rst = 1'b0;
    end

    always #20 clk = ~clk;

    // Instantiate the tester
    tester tester0();

    // Instantiate the drivers
    wishbone_tasks wb0(.*);

    // Instantiate the DUT
    jpeg_top dut(.*);
endmodule // jpeg_top_tb
```

# Testbench: Hi-level tester

```
program tester();  
  int result = 0;  
  int d = 32'h01020304;
```

```
  initial begin
```

```
    for (int i=0; i<16; i++) begin  
      jpeg_top_tb.wb0.m_write(32'h96000000 + 4*i, d); // fill inmem  
      d += 32'h04040404;  
    end
```

```
      jpeg_top_tb.wb0.m_write(32'h96001000, 32'h01000000); // start ax
```

```
      while (result != 32'h80000000)  
        jpeg_top_tb.wb0.m_read(32'h96001000,result); // wait for ax
```

```
      for (int j=0; j<8; j++) begin  
        for (int i=0; i<4; i++) begin // print outmem
```

```
          jpeg_top_tb.wb0.m_read(32'h96000800 + 4*i + j*16,result);
```

```
          $fwrite(1,"%5d ", result >>> 16);
```

```
          $fwrite(1,"%5d ", (result << 16) >>>16);
```

```
        end
```

```
        $fwrite(1,"\n");
```

```
      end
```

```
    end
```

```
  endprogram // tester
```

# mem

```
module mem(wishbone.slave wbm);
  logic [7:0] rom[0:2047];
  logic [1:0] state;
  logic [8:0] adr;
  integer    blockx, blocky, x, y, i;

  initial begin
    // A test image, same as dma_dct_hw.c
    for (blocky=0; blocky<`HEIGHT; blocky++)
      for (blockx=0; blockx<`WIDTH; blockx++)
        for (i=1, y=0; y<8; y++)
          for (x=0; x<8; x++)
            rom[blockx*8+x+(blocky*8+y)*`PITCH] = i++; // these are not wishbone cycles
  end

  assign wbm.err = 1'b0;
  assign wbm.rty = 1'b0;

  always_ff @(posedge wbm.clk)
    if (wbm.rst)
      state <= 2'h0;
    else
      case (state)
        2'h0: if (wbm.stb) state <= 2'h1;
        2'h1: state <= 2'h2;
        2'h2: state <= 2'h0;
      endcase

  assign wbm.ack = state[1];

  always_ff @(posedge wbm.clk)
    adr <= wbm.adr[8:0];

  assign wbm.dat_i = {rom[adr], rom[adr+1], rom[adr+2], rom[adr+3]};
endmodule // mem
```

# DMA? Easy!

```
...
// Init DMA-engine
jpeg_top_tb.wb0.m_write(32'h96001800, 32'h0);
jpeg_top_tb.wb0.m_write(32'h96001804, ?);
jpeg_top_tb.wb0.m_write(32'h96001808, ?);
jpeg_top_tb.wb0.m_write(32'h9600180c, ?);
jpeg_top_tb.wb0.m_write(32'h96001810, ?); // start DMA engine

for (int blocky=0; blocky<`HEIGHT; blocky++) begin
for (int blockx=0; blockx<`WIDTH; blockx++) begin
    // Wait for DCTDMA to fill the DCT accelerator
    result = 0;
    while (?) // wait for block to finish
        jpeg_top_tb.wb0.m_read(32'h96001810, result);

    $display("blocky=%5d blockx=%5d", blocky, blockx);

    for (int j=0; j<8; j++) begin
        for (int i=0; i<4; i++) begin
            jpeg_top_tb.wb0.m_read(32'h96000800 + 4*i + j*16, result);
            $fwrite(1,"%5d ", result >>> 16);
            $fwrite(1,"%5d ", (result << 16) >>>16);
        end
        $fwrite(1,"\n");
    end
end

jpeg_top_tb.wb0.m_write(?); // start next block
end
end
...
```

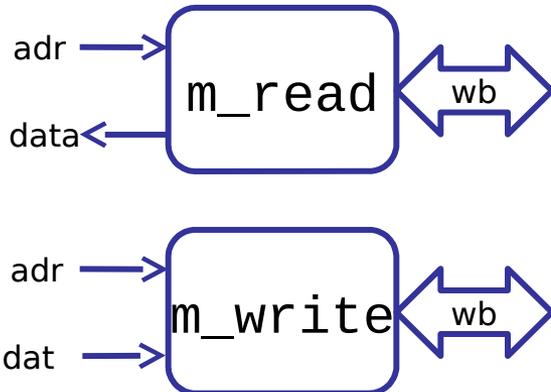
# wishbone\_tasks.sv

- May/may not consume time
- May/may not be synthable
- Do not contain **always/initial**
- Do not return values. Pass via output

```
module wishbone_tasks(wishbone.master wb);
    int result = 0;

    reg oldack;
    reg [31:0] olddat;

    always_ff @(posedge wb.clk) begin
        oldack <= wb.ack;
        olddat <= wb.dat_i;
    end
```



```
// *****
task m_read(input [31:0] adr,
            output logic [31:0] data);

    begin
        @(posedge wb.clk);
        wb.adr <= adr;
        wb.stb <= 1'b1;
        wb.we <= 1'b0;
        wb.cyc <= 1'b1;
        wb.sel <= 4'hf;

        @(posedge wb.clk);
        #1;

        while (!oldack) begin
            @(posedge wb.clk);
            #1;
        end

        wb.stb <= 1'b0;
        wb.we <= 1'b0;
        wb.cyc <= 1'b0;
        wb.sel <= 4'h0;

        data = olddat;
    end
endtask // m_read

// *****
task m_write(input [31:0] adr,
             input [31:0] dat);
    // similar to m_read
endtask // m_write

endmodule // wishbone_tasks
```

# program block

- Purpose: Identifies verification code
- A **program** is different from a **module**
  - only initial blocks allowed
  - executes last  
(module -> clocking/assertions -> program)  
no race situation in previous example!

**The Program block functions pretty much like a C program  
Testbenches are more like software than hardware**

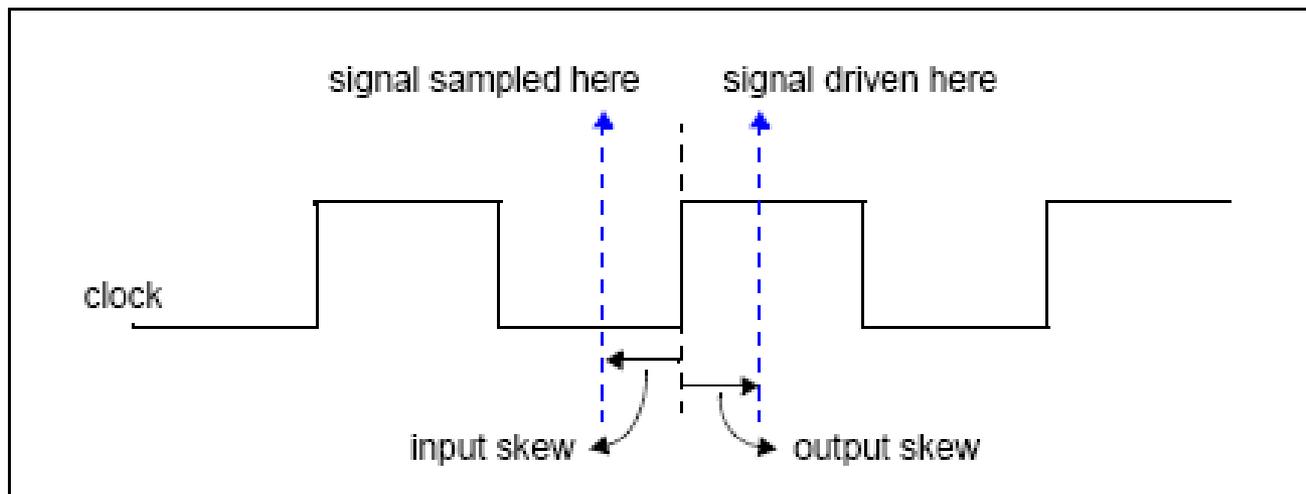
# clocking block

SystemVerilog adds the **clocking block that identifies clock signals, and captures the timing and synchronization requirements** of the blocks being modeled.

A clocking block assembles signals that are synchronous to a particular clock, and makes their timing explicit.

The clocking block is a key element in a cycle-based methodology, which enables users to write testbenches at a higher level of abstraction. Rather than focusing on signals and transitions in time, the test can be defined in terms of cycles and transactions.

## Possible to simulate setup and hold time



# clocking block

```
interface wishbone(input clk,rst);
    wire stb,ack;

    clocking cb @(posedge clk);
        input ack;
        output stb;
    endclocking // cb

    modport tb (clocking cb,
                input clk,rst);

endinterface // wishbone
```

```
module tb();
    logic        clk = 1'b0;
    logic        rst = 1'b1;

    // instantiate a WB
    wishbone wb(clk,rst);

    initial begin
        #75 rst = 1'b0;
    end

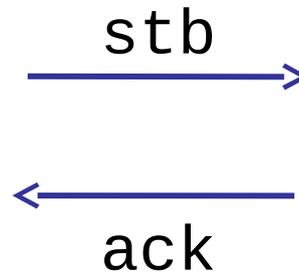
    always #20 clk = ~clk;

    // Instantiate the DUT
    jpeg_top dut(.);

    // Instantiate the tester
    tester tester0(.);
endmodule // jpeg_top_tb
```

# clocking block

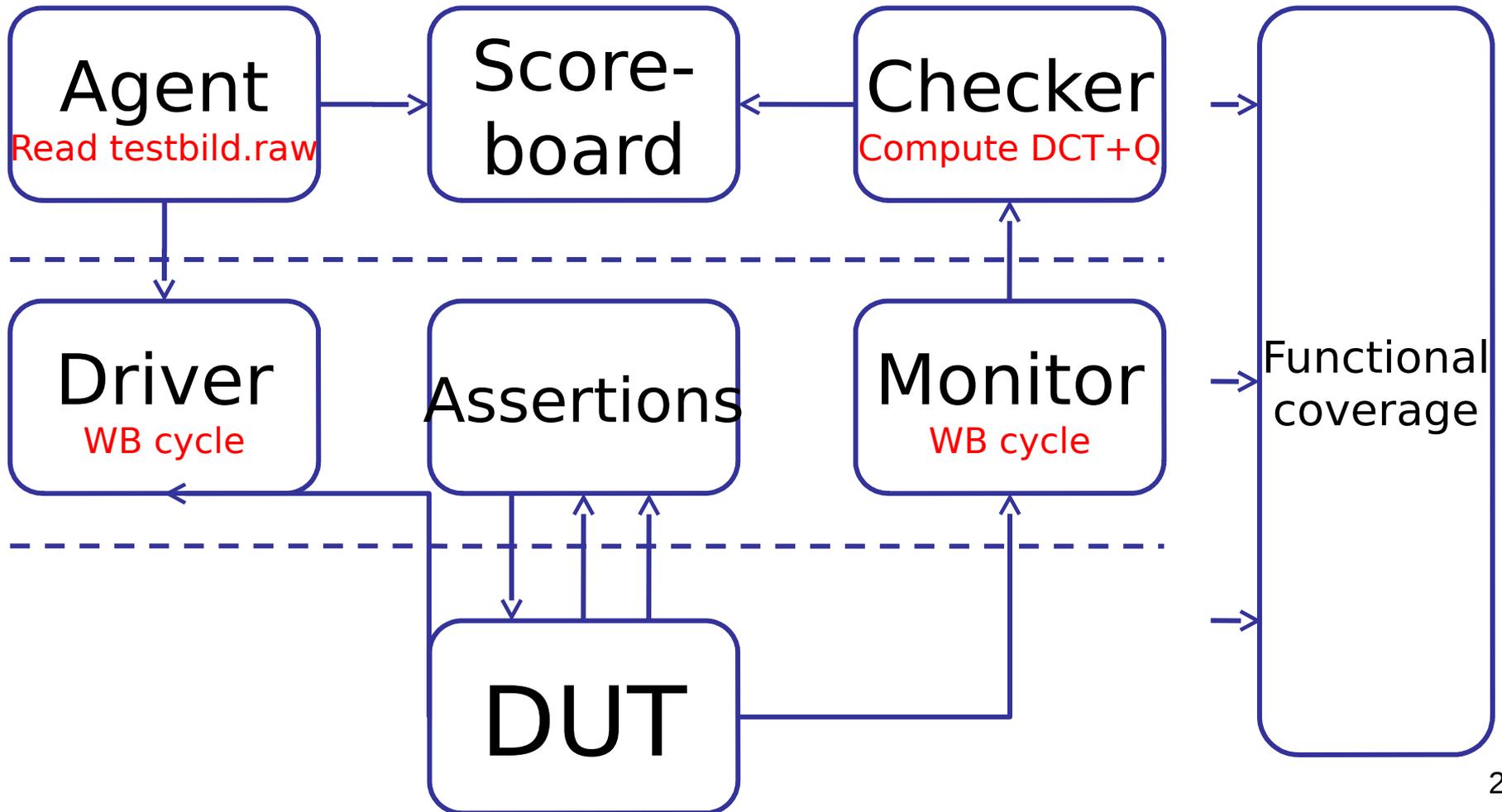
```
program tester(wishbone.tb wb);  
  
...  
  
initial begin  
    for (int i=0; i<3; i++) begin  
        wb.cb.stb <= 0;  
        ##1;  
        wb.cb.stb <= 1;  
        while (wb.cb.ack==0)  
            ##1;  
        end  
    end  
endprogram // tester
```



```
module jpeg_top(wishbone wb);  
    reg state;  
  
    assign wb.ack = state;  
  
    always_ff @(posedge wb.clk)  
        if (wb.rst)  
            state <= 1'b0;  
        else if(state)  
            state <= 1'b0;  
        else if (wb.stb)  
            state <= 1'b1;  
    endmodule // jpeg_top
```

# A complex testbench

from Spear: SV for verification



# Object oriented Programming

- SV includes OOP
- Classes can be defined
  - inside a program
  - inside a module
  - stand alone

# OOP

```
program class_t;

class packet;
// members in class
integer size;
integer payload [];
integer i;
// Constructor
function new (integer size);
begin
    this.size = size;
    payload = new[size];
    for (i=0; i < this.size; i ++)
        payload[i] = $random();
end
endfunction
// Task in class (object method)
task print ();
begin
    $write("Payload : ");
    for (i=0; i < size; i ++)
        $write("%x ", payload[i]);
    $write("\n");
end
endtask
```

```
// Function in class (object method)
function integer get_size();
begin
    get_size = this.size;
end
endfunction
endclass

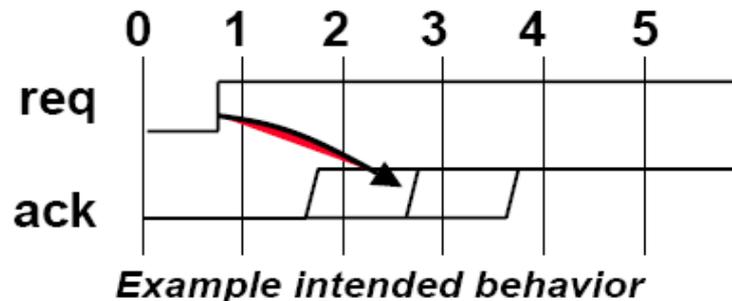
packet pkt;

initial begin
    pkt = new(5);
    pkt.print();
    $display ("Size of packet %0d",
              pkt.get_size());
end

endprogram
```

# What is an assertion?

- A concise description of [un]desired behavior



**“After the request signal is asserted, the acknowledge signal must come 1 to 3 cycles later”**

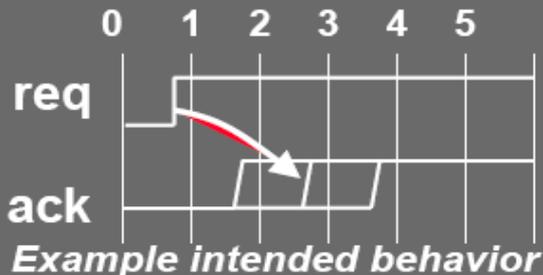
# Assertions

SVA Assertion

```
property req_ack;  
  @(posedge clk) req ##[1:3] $rose(ack);  
endproperty  
as_req_ack: assert property (req_ack);
```

VHDL

```
sample_inputs : process (clk)  
begin  
  if rising_edge(clk) then  
    STROBE_REQ <= REQ;  
    STROBE_ACK <= ACK;  
  end if;  
end process;  
protocol: process  
  variable CYCLE_CNT : Natural;  
begin  
  loop  
    wait until rising_edge(CLK);  
    exit when (STROBE_REQ = '0') and (REQ = '1');  
  end loop;  
  CYCLE_CNT := 0;  
  loop  
    wait until rising_edge(CLK);  
    CYCLE_CNT := CYCLE_CNT + 1;  
    exit when ((STROBE_ACK = '0') and (ACK = '1')) or (CYCLE_CNT = 3);  
  end loop;  
  if ((STROBE_ACK = '0') and (ACK = '1')) then  
    report "Assertion success" severity Note;  
  else  
    report "Assertion failure" severity Error;  
  end if;  
end process protocol;
```



HDL Assertion

# Assertions

Assertions are built of

- 
1. Boolean expressions
  2. Sequences
  3. Properties
  4. Assertion directives

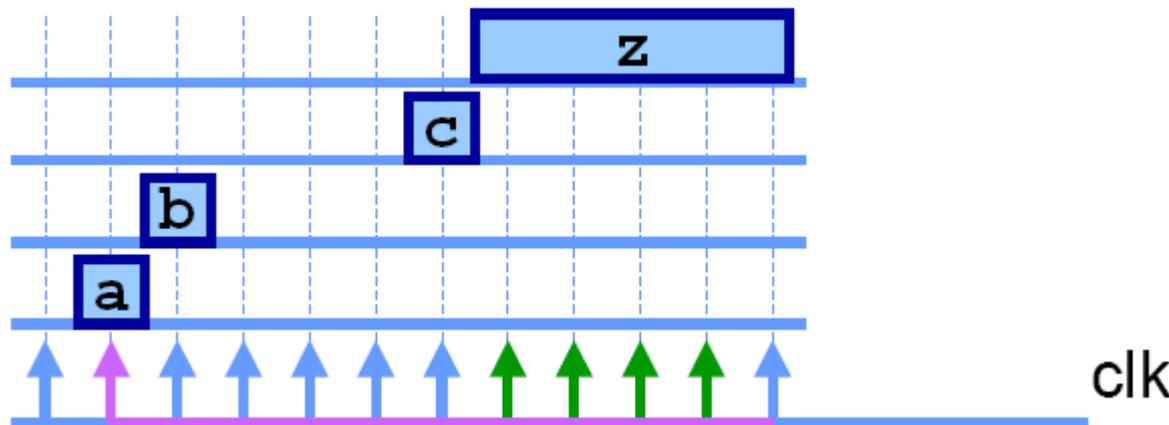
# Sequential regular expressions

- Describing a sequence of events
- Sequences of Boolean expressions can be described with a specified time step in-between
- ##N delay operator
- [\*N] repetition operator

**sequence s1;**

```
@(posedge clk) a ##1 b ##4 c ##[1:5] z;
```

**endsequence**



- signal
- expression
- sequence

# Coverage

- *Code Coverage (code profiling)*  
reflects how thorough the HDL code was exercised.
- *Functional Coverage (histogram binning)*  
perceives the design from a user's or a system point of view.  
Have you covered all of your typical scenarios?  
Error cases? Corner cases? Protocols?
- Functional coverage also allows relationships,  
"OK, I've covered every state in my state machine, but did I ever have an interrupt at the same time? When the input buffer was full, did I have all types of packets injected? Did I ever inject two erroneous packets in a row?"

# Coverage

```
// DUT With Coverage
module simple_coverage();

logic [7:0]  addr;
logic [7:0]  data;
logic       par;
logic       rw;
logic       en;

// Coverage Group
covergroup memory @ (posedge en);
  address : coverpoint addr {
    bins low    = {0,50};
    bins med    = {51,150};
    bins high   = {151,255};
  }
  parity : coverpoint par {
    bins even   = {0};
    bins odd    = {1};
  }
  read_write : coverpoint rw {
    bins read   = {0};
    bins write  = {1};
  }
endgroup
```

```
memory mem = new();
```

```
// Task to drive values
task drive (input [7:0] a, input [7:0] d,
           input r);
    #5 en <= 1;
    addr <= a;
    rw <= r;
    data <= d;
    par <= ^d;
    $display ("%2tns Address :%d data %x,
              rw %X, parity %X",
              $time,a,d,r, ^d);
    #5 en <= 0;
    rw <= 0;
    data <= 0;
    par <= 0;
    addr <= 0;
    rw <= 0;
endtask

// Testvector generation
initial begin
  en = 0;
  repeat (10) begin
    drive ($random,$random,$random);
  end
  #10 $finish;
end

endmodule
```

# Report

ModelSim says:

```
# @ 5ns Address : 36 data 81, rw 1, parity 0
# @15ns Address : 99 data 0d, rw 1, parity 1
# @25ns Address :101 data 12, rw 1, parity
0
# @35ns Address : 13 data 76, rw 1, parity 1
# @45ns Address :237 data 8c, rw 1, parity
1
# @55ns Address :198 data c5, rw 0, parity
0
# @65ns Address :229 data 77, rw 0, parity
0
# @75ns Address :143 data f2, rw 0, parity 1
```

COVERGROUP COVERAGE:

```
# @85ns Address :232 data c5, rw 0, parity
0
# @95ns Address :189 data 2d, rw 1, parity
0
```

Covergroup	Metric	Goal/ Status
	At Least	
TYPE /simple_coverage/memory		44.4% 100 Uncovered
Coverpoint memory::address		33.3% 100 Uncovered
covered/total bins:	1	3
bin low	9	1 Covered
bin med	0	1 ZERO
bin high	0	1 ZERO
Coverpoint memory::parity		50.0% 100 Uncovered
covered/total bins:	1	2
bin even	9	1 Covered
bin odd	0	1 ZERO
Coverpoint memory::read_write		50.0% 100 Uncovered
covered/total bins:	1	2
bin read	9	1 Covered
bin write	0	1 ZERO

Report generator:

TOTAL COVERGROUP COVERAGE: 44.4% COVERGROUP TYPES: 1

# Cross coverage

```
enum { red, green, blue } color;
bit [3:0] pixel_adr;

covergroup g1 @(posedge clk);
  c: coverpoint color;
  a: coverpoint pixel_adr;
  AxC: cross color, pixel_adr;
endgroup;
```

Sample event

3 bins for color

16 bins for pixel

48 (=16 \* 3)  
cross products

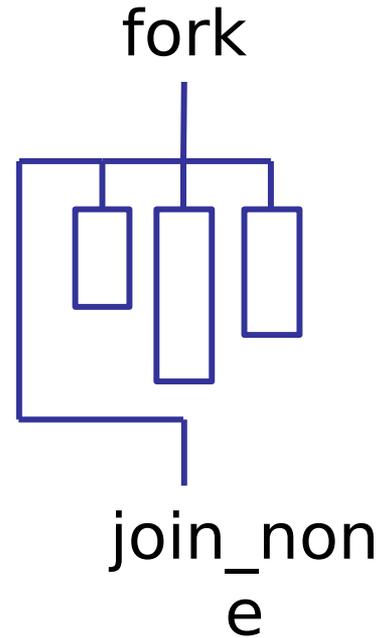
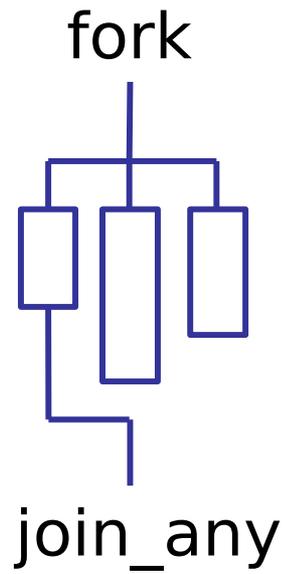
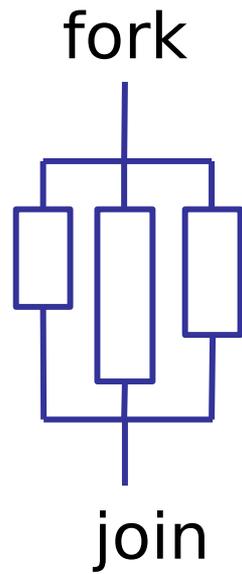
# Constrained randomization

```
program rc;

class Bus;
    rand bit[31:0] addr;
    rand bit[31:0] data;
    constraint word_align {addr[1:0] == 2'b0;
                           addr[31:24] == 8'h99;}
endclass // Bus

initial begin
    Bus bus = new;
    repeat (50) begin
        if ( bus.randomize() == 1 )
            $display ("addr = 0x%h    data = 0x%h\n",
                     bus.addr, bus.data);
        else
            $display ("Randomization failed.\n");
        end
    end
endprogram // rc
```

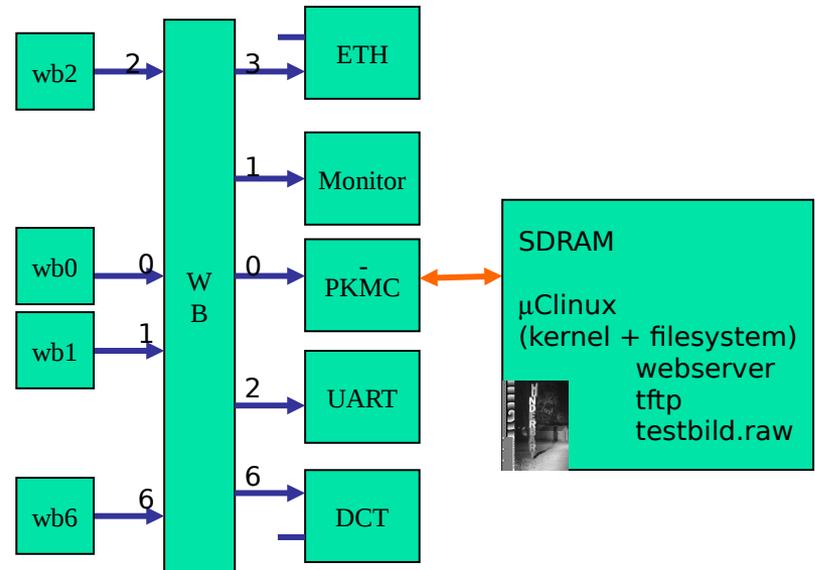
# Parallel threads



# An example- sketch

## WB arbitration test

instantiate 4 wishbone\_tasks  
(must be declared automatic)



```
program tester2();  
...  
initial begin  
...  
fork  
begin // 2  
for (int i; i<100; i++)  
jpeg_top_tb.wb2.m_write(32'h100, 32'h0);  
end  
...  
begin // 6  
for (int i; i<100; i++)  
jpeg_top_tb.wb6.m_write(32'h20000000, result);  
end  
...  
join  
...  
end  
endprogram
```