

Lecture 3, ANIK

Current mirrors, Improved gain stages

diléda mante

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Mea sentencia

What did we do last time?

"Simple" amplifier stages

A single transistor can be troublesome enough ...

Small-signal schematics practice

Practice, practice, practice

How do we increase gain?

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What are our handles?

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What will we do today?

Swing

How many transistors can we stack?

Improving the gain

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What are our handles? (again)

We need high gain - how do we do it?

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Voltage swing

Walk around the circuit

Check for all the required voltage levels to maintain transistors in their saturation region

Use the following relations

$$V_{GS} = V_{EFF} + V_T$$
, $V_{DS} > V_{EFF} \Rightarrow V_{DS} = V_{EFF}$, $V_{EFF} = \sqrt{\frac{I_D}{\alpha}}$

The lower v_{eff} ...

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the higher swing

the higher gain

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Examples on the board

Consider the three amplifiers

CG, CD, CS

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Check the potentials

When are the transistors saturated?

What does this imply?



How do I increase my gain?

Assuming a simple common-source stage:

$$A = \frac{g_m}{g_{out}} = \frac{1}{\lambda \cdot v_{eff}} = \frac{2\sqrt{\alpha}}{\lambda \sqrt{I_D}}$$

The answer depends on the biasing conditions

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Decrease V_{eff}
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Decrease $\lambda \sim 1/L$, i.e., increase the channel length.

Decrease (!) the current I_D

Increase the transistor sizes, $\alpha \sim S \sim W$



Another observation

DC gain vs the effective

voltage and currents



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Gain is formed by simply taking the product of all gains

Offers high swing

Might cost us more power consumption (each stage needs a current).

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Improving the gain, the electrical option

Revisit the expression on gain!

$$A = \frac{g_m}{g_{out}}$$

Increase the transconductance

Decrease the output conductance (i.e., increase output impedance)

We've done that kind of, c.f., lowering the v_{eff} , etc.

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Cascodes, the hardware option

Introduce more hardware to increase impedance

Cascodes increase the gain

How? - a small-signal exercise

We must balance the load

both in PMOS and NMOS "direction"

(Traditional way to maximize power efficiency)



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So ... it's all about impedance levels

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Cascodes

(Quickly) eats up the voltage headroom

For every diode-connected transistor, we loose

one V_T of swing

We can save current since only one stage

Complexing biasing schemes

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"The output impedance is multiplied"





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Cascodes, common-source example

Voltage swing

Calculating the gain



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Cascodes, common-source example

This formula still holds $A = g_m / g_{out}$ and the output conductance is

$$A = \frac{g_{m1}}{\frac{g_{n1} \cdot g_{n2}}{g_{m2}} + \frac{g_{p3} \cdot g_{p4}}{g_{m3}}} \approx \frac{g_{m1} \cdot g_{m2}}{2 \cdot g_{n1} \cdot g_{n2}}$$

Now, we have some more handles to increase (set) the gain.

Effective voltage of input can be decoupled.

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Classical analog trade-offs to distribute the gain).

But ... what happens to the gain if the impedance levels are not balanced?

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Some conlusions on one slide



Cascodes eat up the swing

Cascodes save current compared to multi-stage

Cascodes and multi-stage have comparatively same area

Cascodes have more complex biasing schemes compared to multi-stage

Cascodes might not be feasible in future (analog) designs



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What did we do today?

Voltage swing

Cascodes to increase gain

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What will we do next time?

Current mirrors

Simple, Wilson, Wide-Swing, Cascoded

Improved amplifier stages

Folded-cascoded gain stage

Decoupling design parameters

Using for example current mirrors



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