

TSTE19 Power Electronics

- Lecture 15
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 - ISY/EKS

Outline

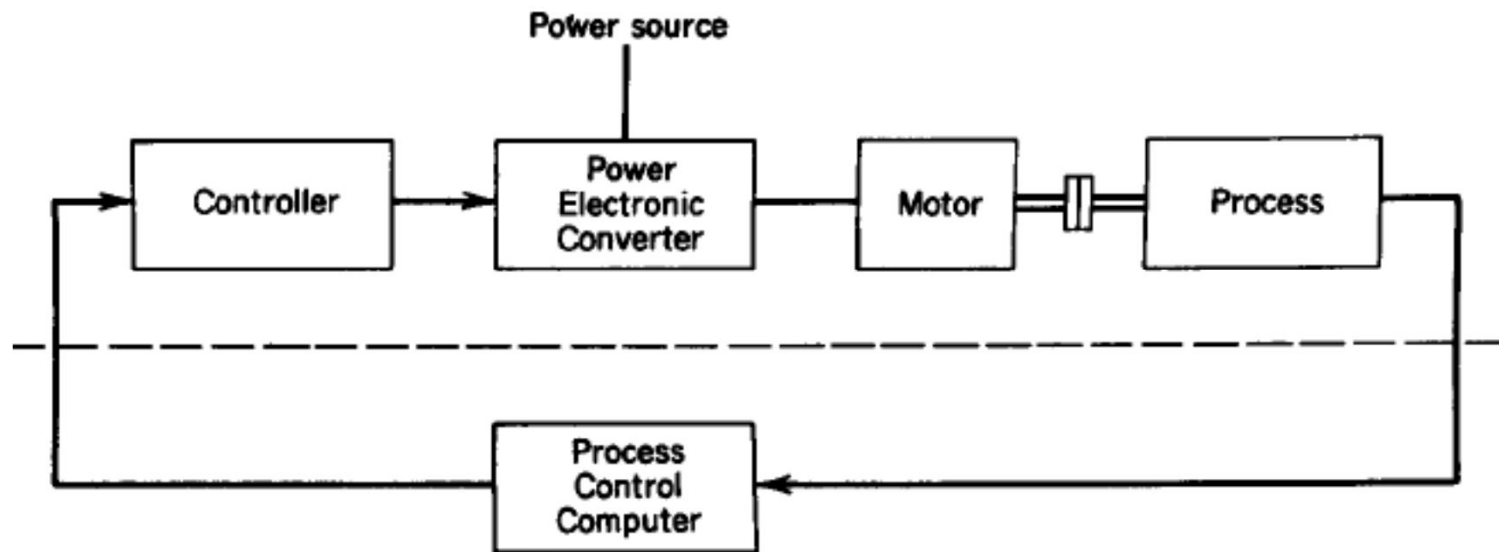
- Motor drives
- Harmonics

Motor drives

- Details on electrical motor principles, design, and control found in other course
 - TSFS04 Electrical Drives
- Motor control is a common application of power electronics
- Three major types
 - DC motors
 - Induction motors
 - Synchronous motors

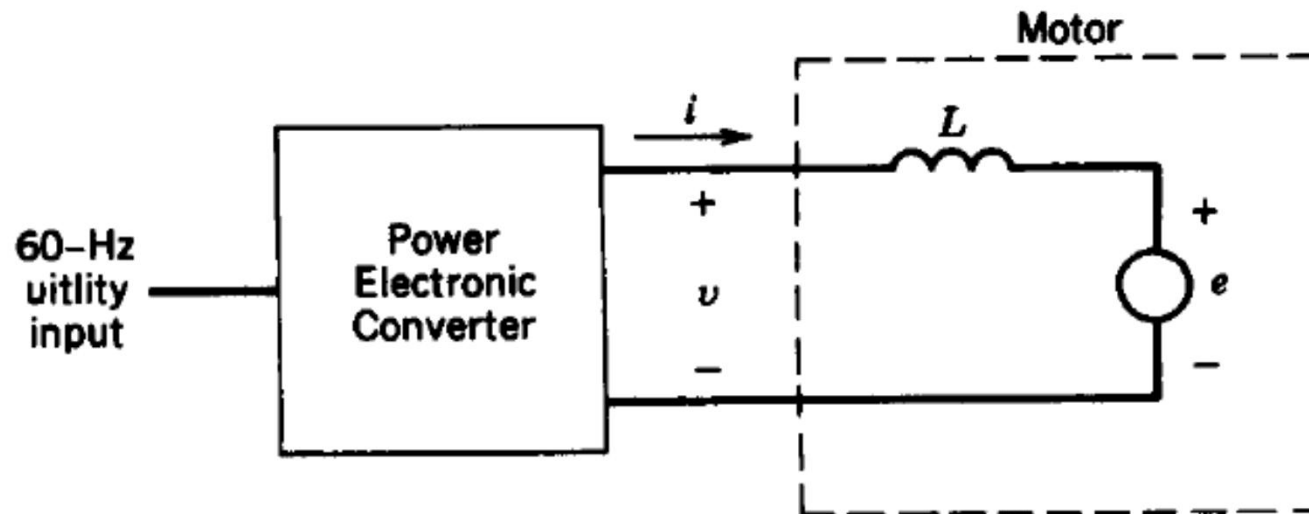
Electric motor control structure

- Process may return speed and/or position as well as temperature or other sensor values
 - Motor speed are sometimes not interesting, e.g. in ventilation systems



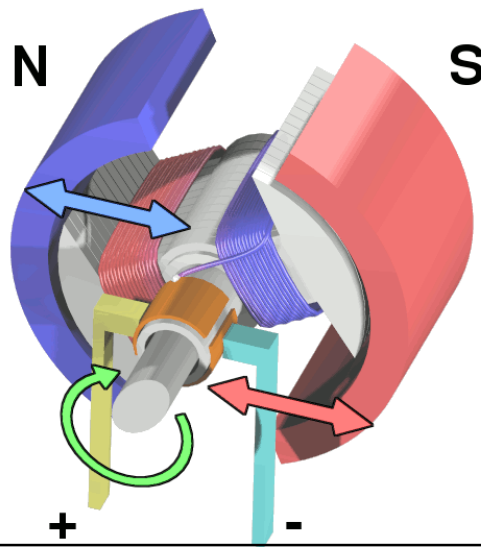
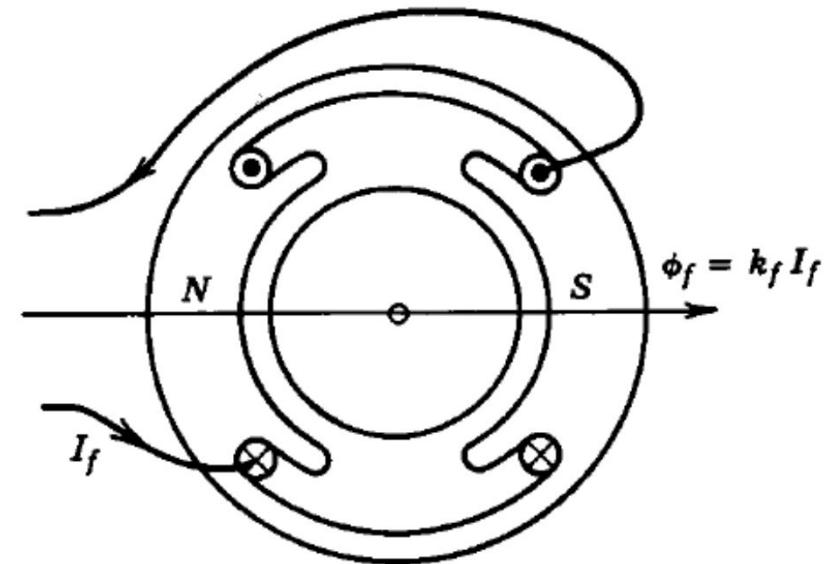
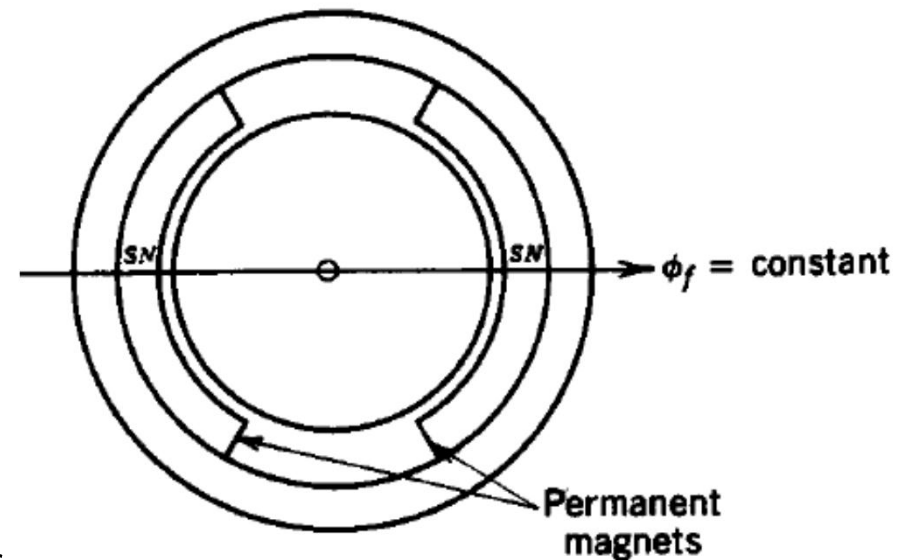
Simplified circuit of a drive

- Counter-emf e increase linearly with motor speed
- Voltage rating of power electronic converter defined by maximum motor speed
- Current rating defined by the mechanical torque



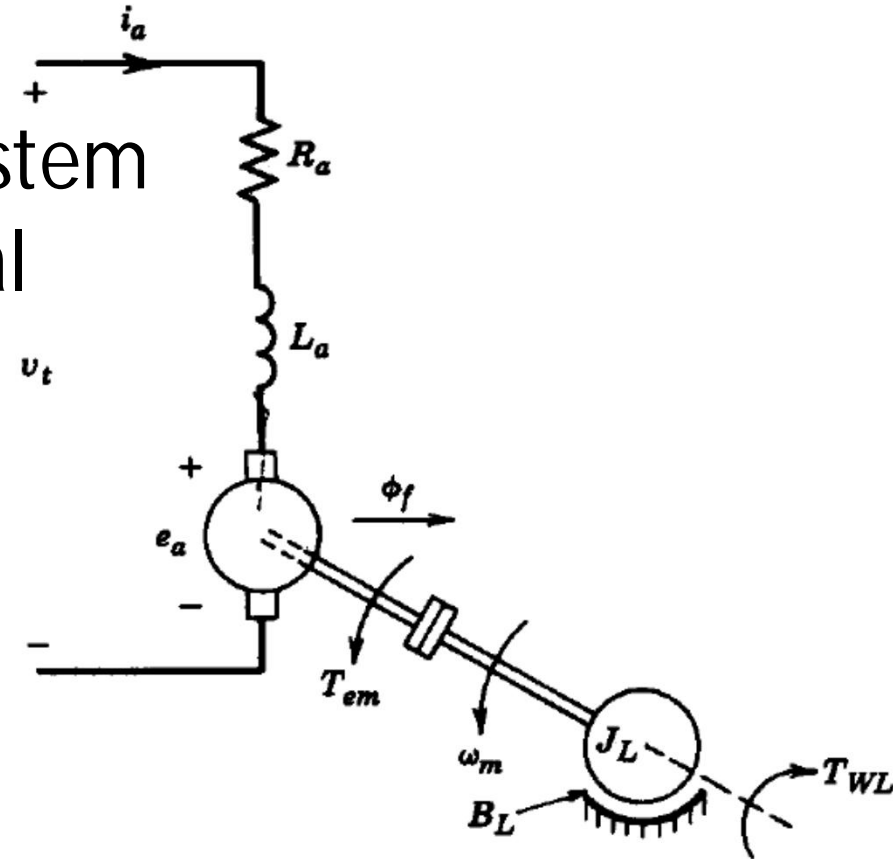
DC-Motor drive

- Stator field by permanent magnets or field windings
- Rotor contains armature winding
 - Handles the electrical power
 - Connected through carbon brushes



DC motor equivalent circuit

- Includes mechanical system modeled as an electrical circuit
- T_{em} = Electromagnetic torque
- ω_m = speed
- J, B equivalent inertia and damping



DC-motor equivalent

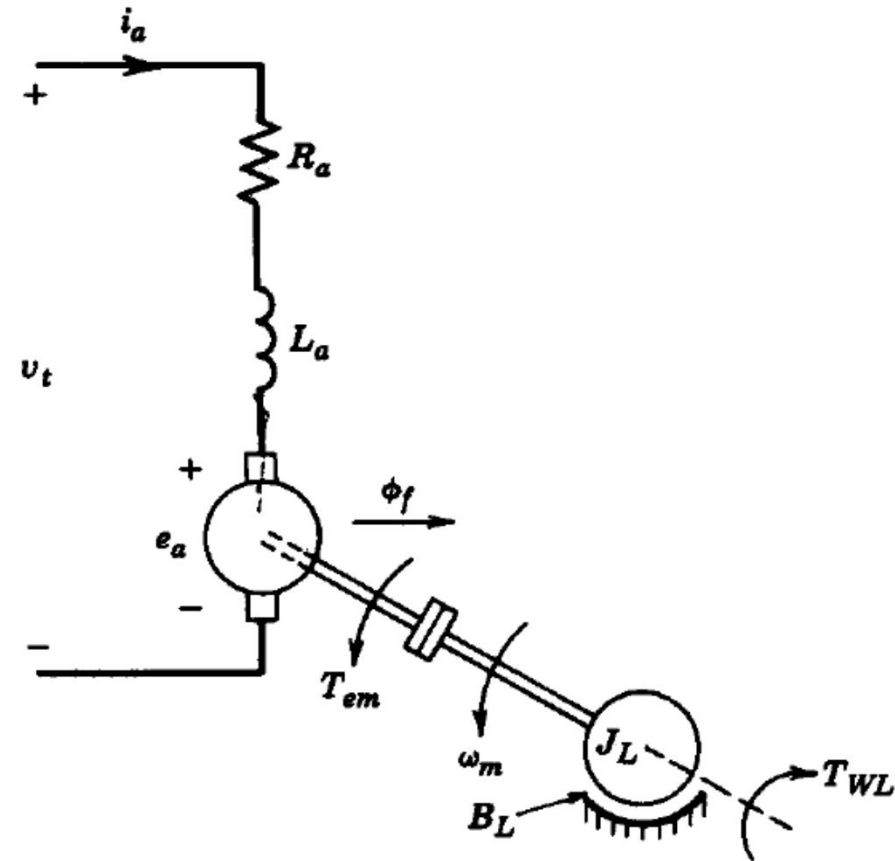
- $T_{em} = k_t \phi_f i_a$,
- $P_m = \omega_m T_{em} = k_t \phi_f \omega_m i_a$
- $e_a = k_e \phi_f \omega_m$
- $P_e = e_a i_a = k_e \phi_f \omega_m i_a$

Steady state:

- $P_e = P_m \Rightarrow k_e = k_t$

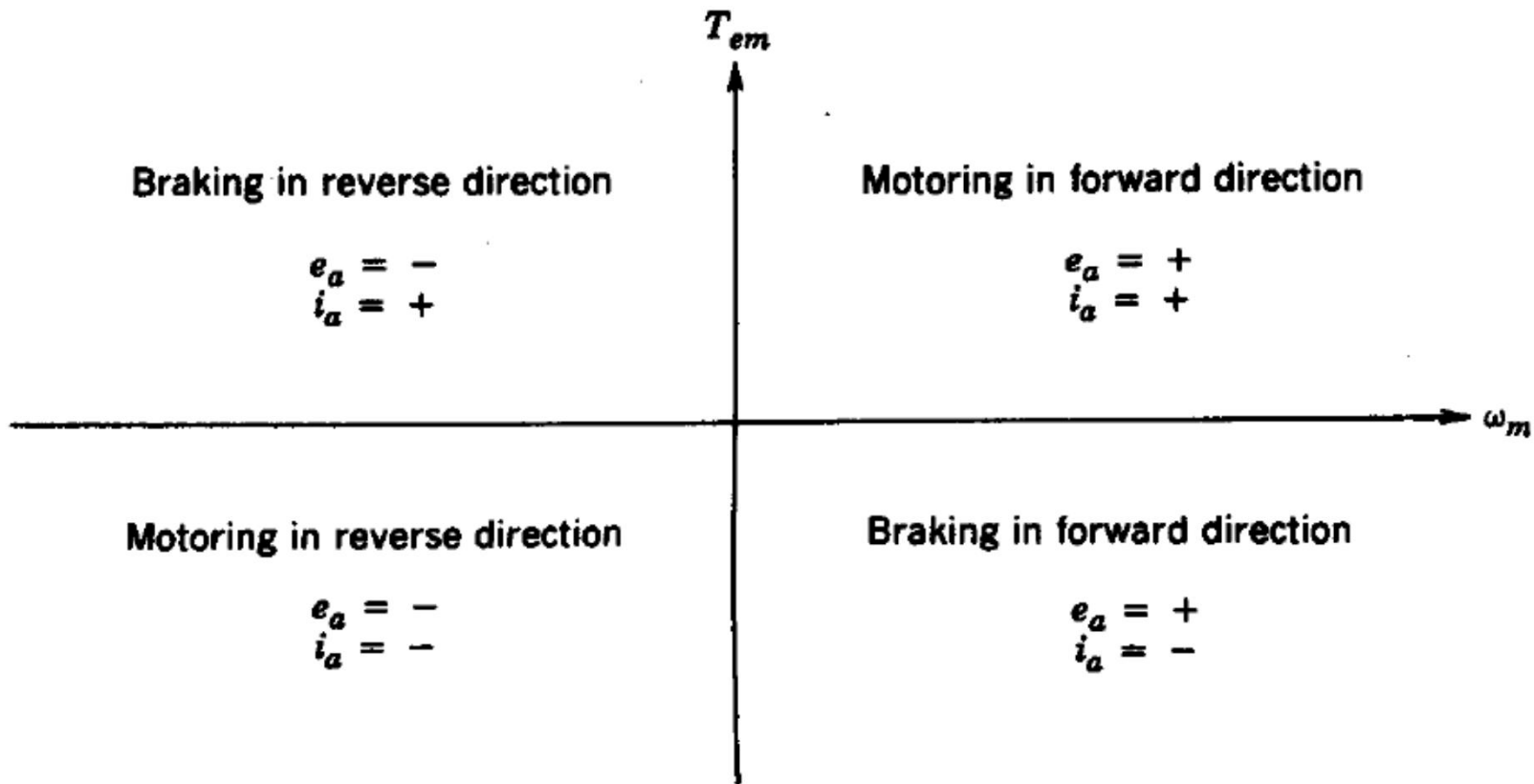
Equivalent circuit

- $v_t = e_a + R_a i_a + L_a \frac{di_a}{dt}$



Operation of DC motor

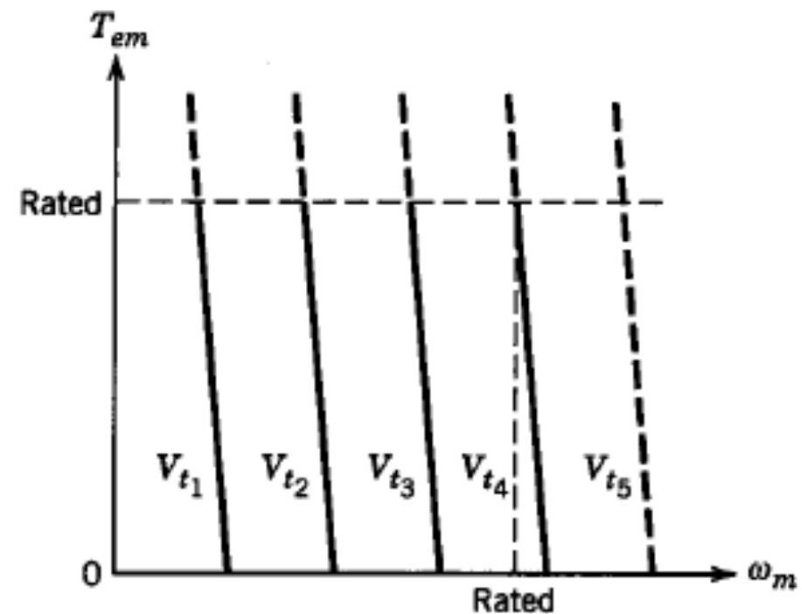
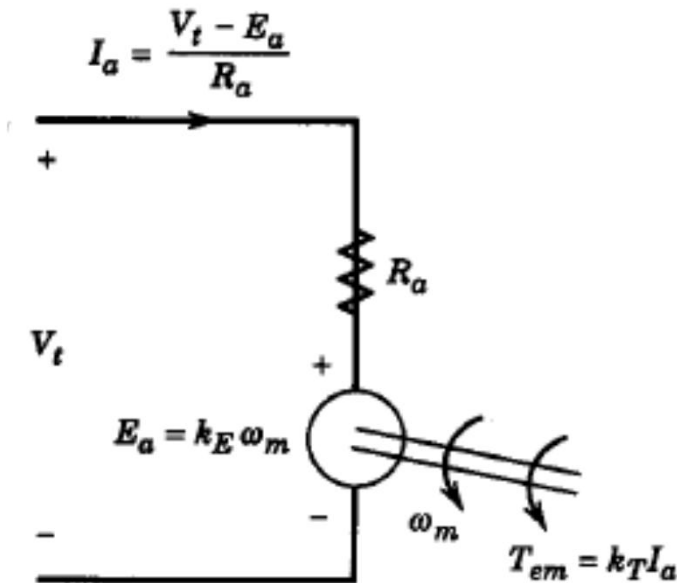
- Can be reversed as well as act as a generator



Control of DC motor

- Control of V_t controls torque T_{em}

- $T_{em} = k_T I_a$
 - $k_T = k_t \phi_f$
- $E_a = k_E \omega_m$
 - $k_E = k_e \phi_f$
- $V_t = E_a + R_a I_a$



Induction motor drive

- No carbon brushes
- Rotating stator magnetic field
 - Requires three or more phases
 - Fixed speed
- Induces a magnetic field in the rotor winding
- Synchronous speed
 - $\omega_s = 2(2 \cdot \pi \cdot f) / p = 2\omega / p$
 - $p =$ number of poles in motor
 - $n_s = 60 \cdot \frac{\omega_s}{2\pi} = \frac{120}{p} f$ [rpm]
- Slip speed $\omega_{sl} = \omega_s - \omega_r$
 - Slip $s = (\omega_s - \omega_r) / \omega_s$

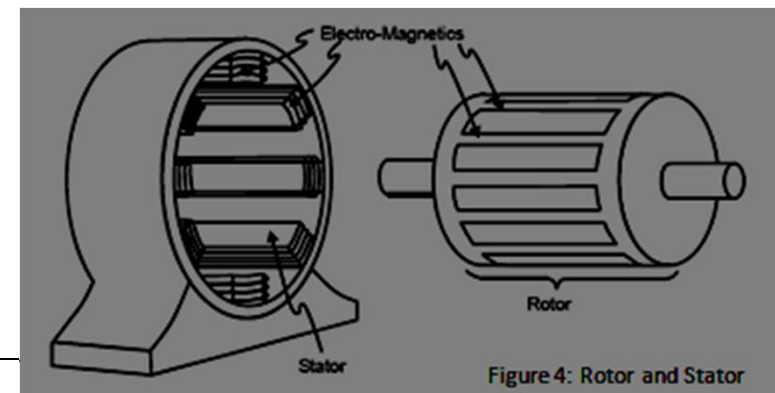
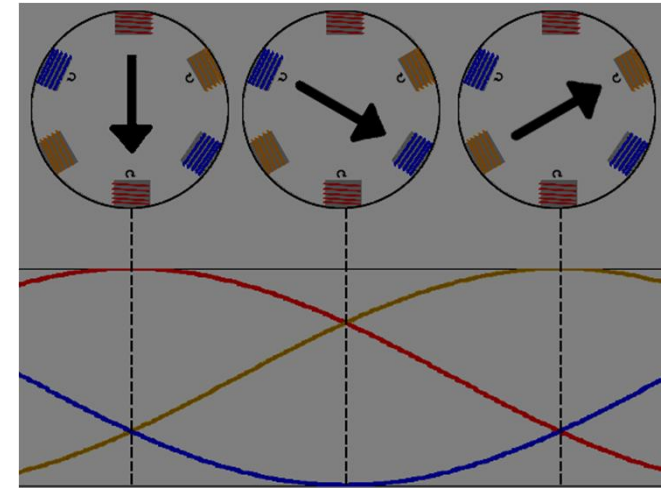
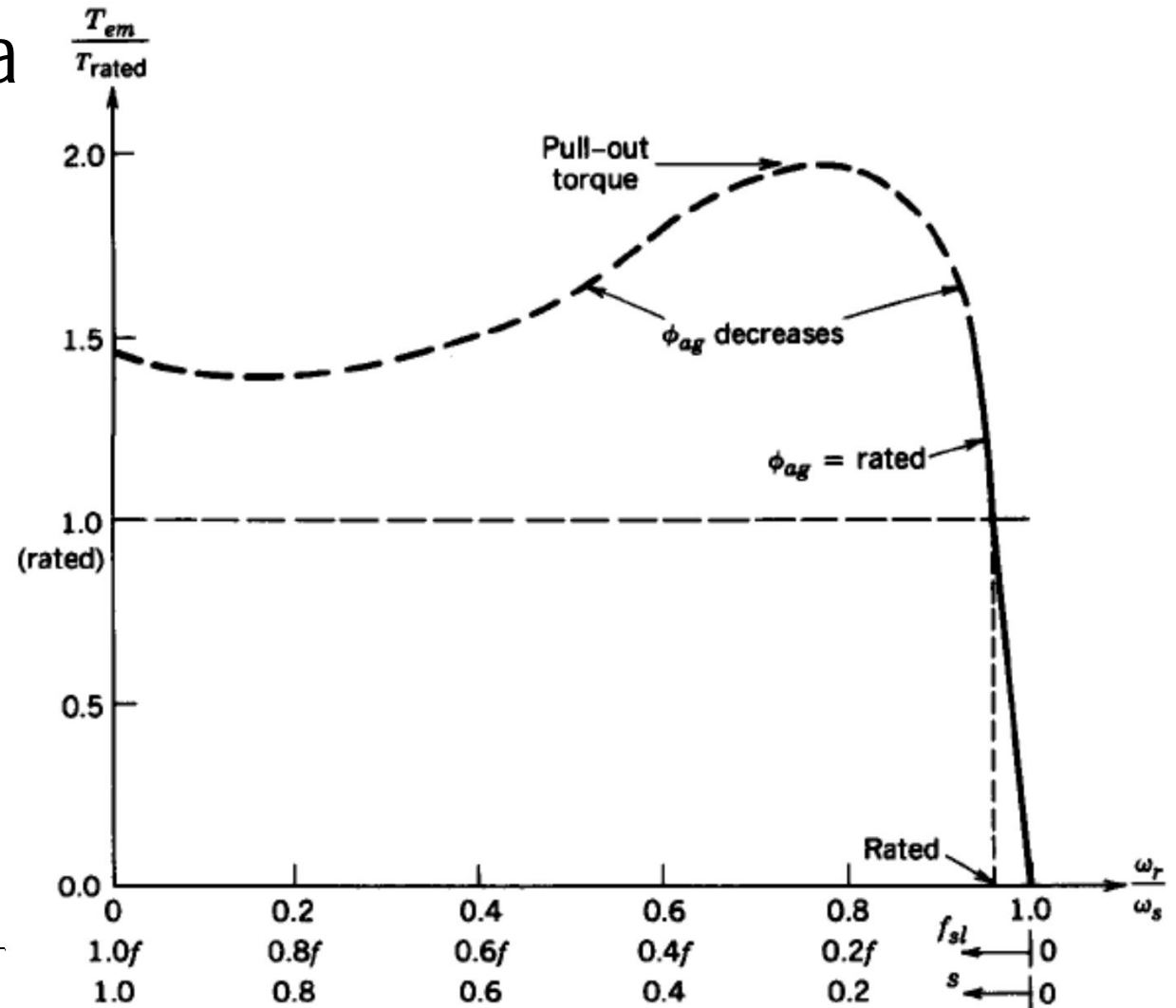


Figure 4: Rotor and Stator

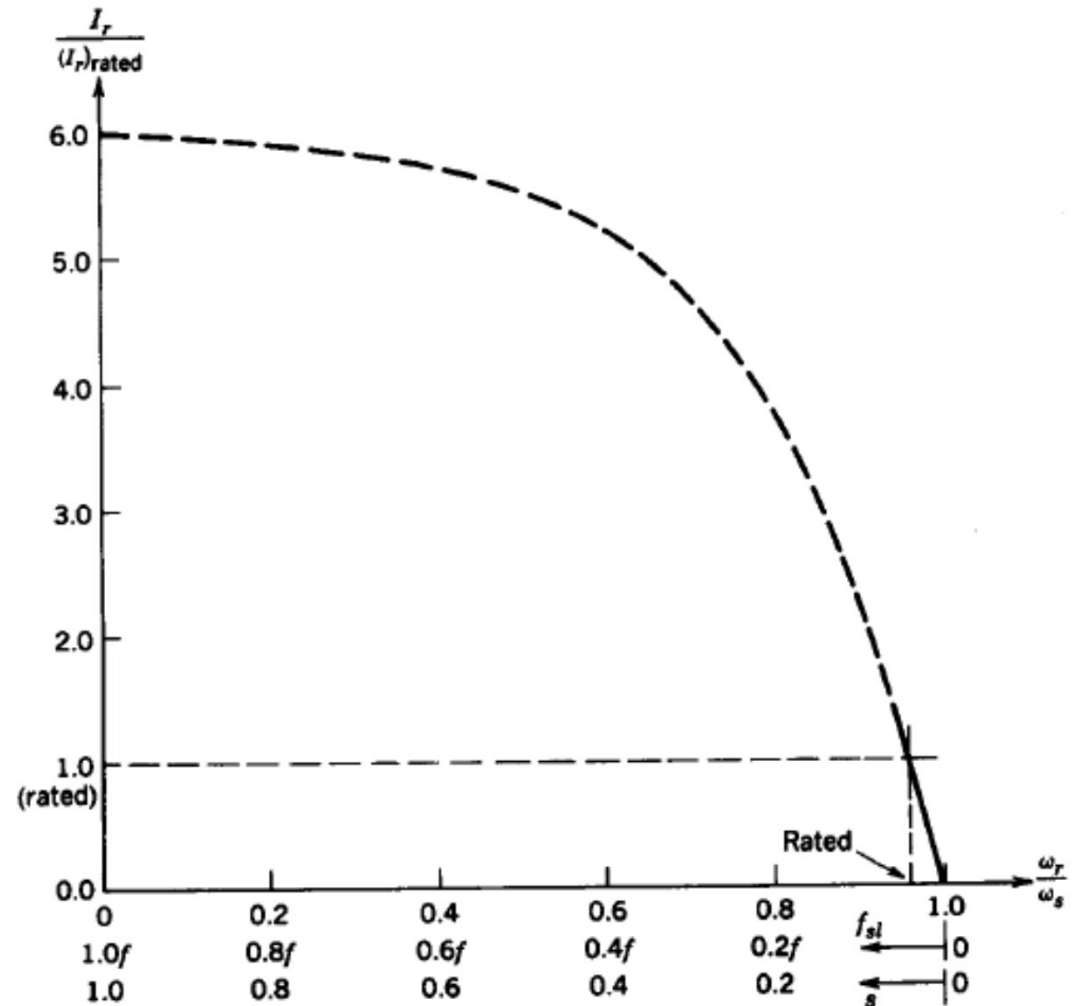
Induction motor

- Linear area used for speed control



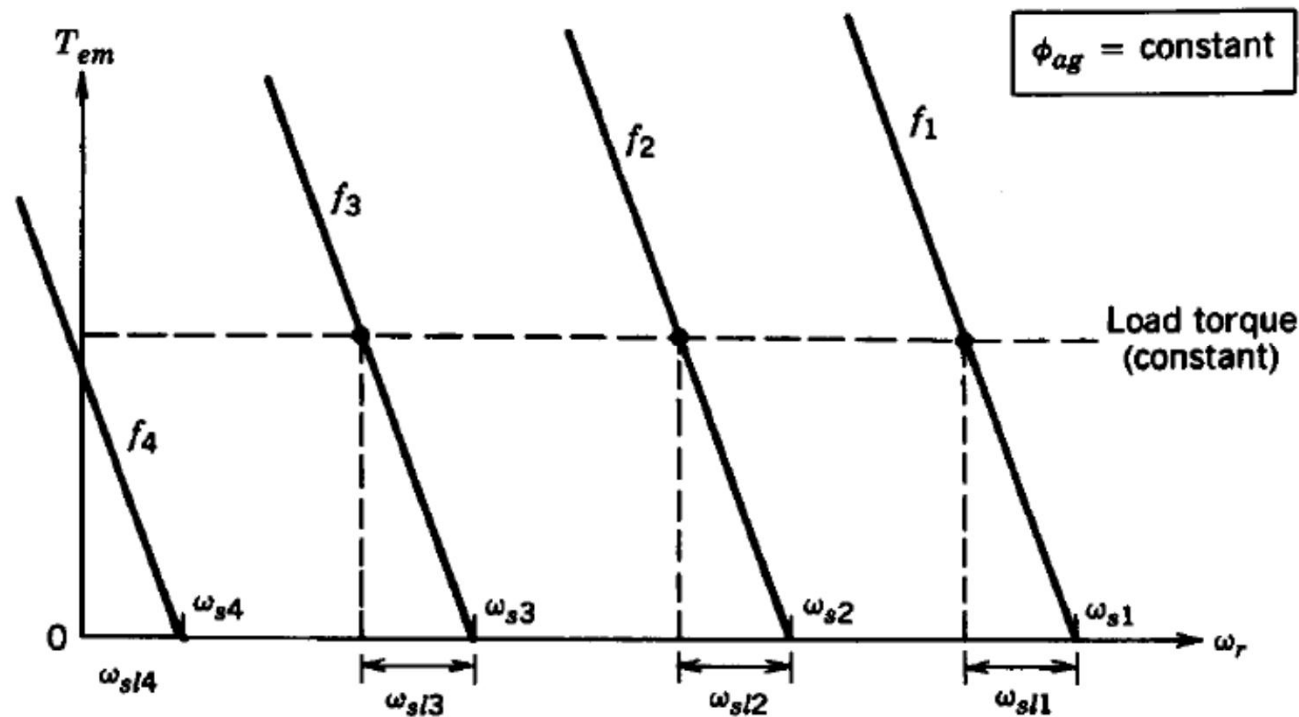
Induction motor current

- Fixed V_s and f
- Large start-up current if not controlled

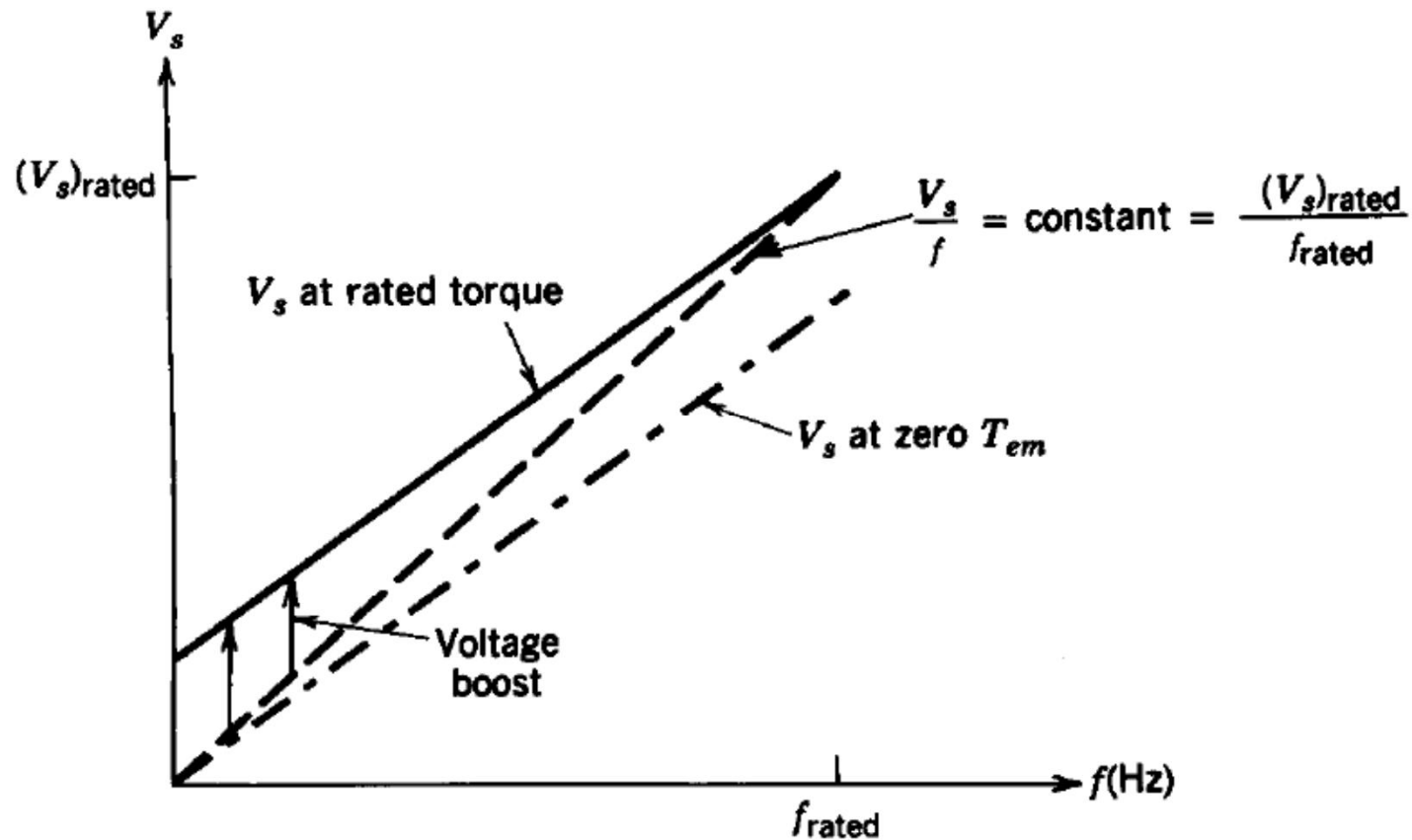


Induction motor speed control

- Air gap flux constant by changing V_s proportional to f

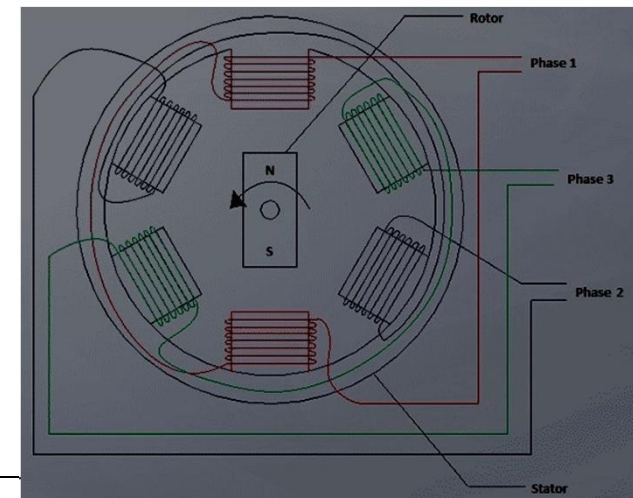


Constant air-gap flux: $\frac{V_s}{f} = \text{constant}$



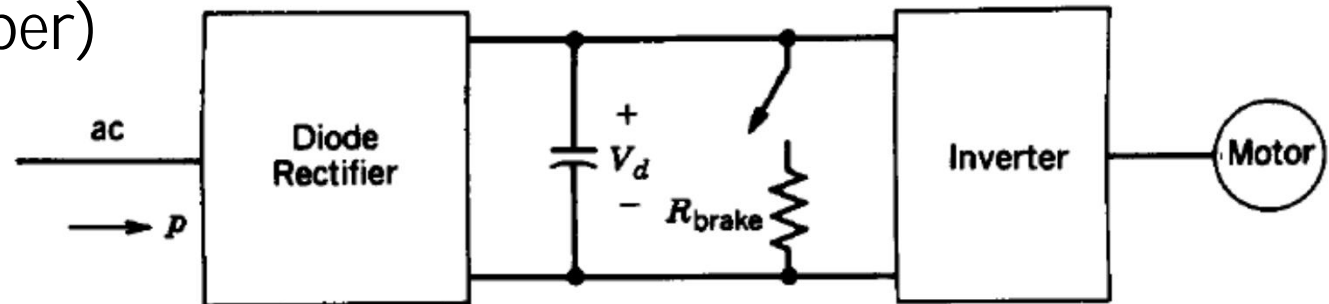
Synchronous motor drive

- Fixed rotor magnetization
 - Permanent magnet rotor winding (larger motors)
- Rotating magnetic field in the field winding
 - Motor speed defined by magnetic rotation speed
- Also known as “brushless dc” motor

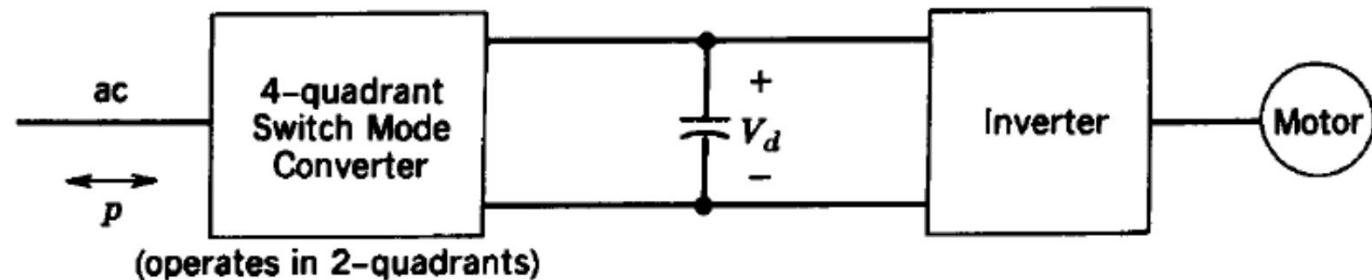


Converter system for motor drive

- PWM-VSI (Voltage Source Inverter)
 - a) Diode rectifier – Dissipative braking (chopper)



- b) 4-quadrant switch mode rectifier – Regenerative braking



Lecture 15

Harmonics

Fourier series

$$f(t) = F_0 + \sum_{h=1}^{\infty} f_h(t) = \frac{1}{2}a_0 + \sum_{h=1}^{\infty} \{a_h \cos(h\omega t) + b_h \sin(h\omega t)\}$$

where $F_0 = \frac{1}{2}a_0$ is the average value

$$a_h = \frac{1}{\pi} \int_0^{2\pi} f(t) \cos(h\omega t) d(\omega t) \quad h = 0, \dots, \infty$$

$$b_h = \frac{1}{\pi} \int_0^{2\pi} f(t) \sin(h\omega t) d(\omega t) \quad h = 1, \dots, \infty$$

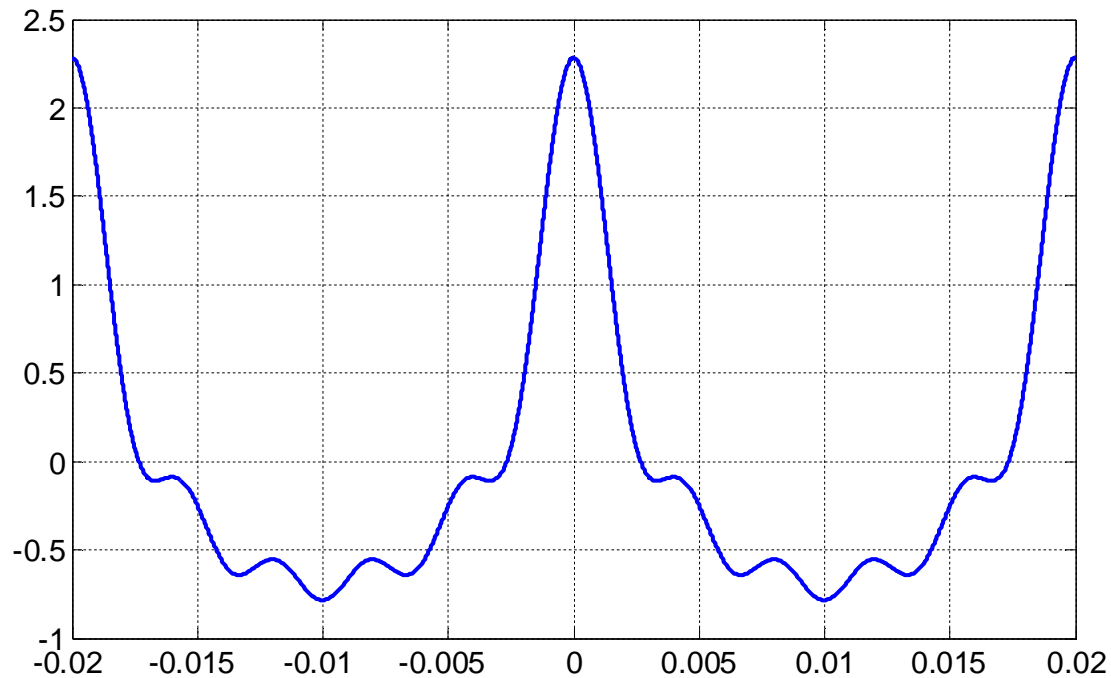
Fourier analysis

Table 3-1 Use of Symmetry in Fourier Analysis

| <i>Symmetry</i> | <i>Condition Required</i> | <i>a_h and b_h</i> |
|-------------------|-------------------------------|--|
| Even | $f(-t) = f(t)$ | $b_h = 0$ $a_h = \frac{2}{\pi} \int_0^{\pi} f(t) \cos(h\omega t) d(\omega t)$ |
| Odd | $f(-t) = -f(t)$ | $a_h = 0$ $b_h = \frac{2}{\pi} \int_0^{\pi} f(t) \sin(h\omega t) d(\omega t)$ |
| Half-wave | $f(t) = -f(t + \frac{1}{2}T)$ | $a_h = b_h = 0$ for even h $a_h = \frac{2}{\pi} \int_0^{\pi} f(t) \cos(h\omega t) d(\omega t)$ for odd h $b_h = \frac{2}{\pi} \int_0^{\pi} f(t) \sin(h\omega t) d(\omega t)$ for odd h |
| Even quarter-wave | Even and half-wave | $b_h = 0$ for all h $a_h = \begin{cases} \frac{4}{\pi} \int_0^{\pi/2} f(t) \cos(h\omega t) d(\omega t) & \text{for odd } h \\ 0 & \text{for even } h \end{cases}$ |
| Odd quarter-wave | Odd and half-wave | $a_h = 0$ for all h $b_h = \begin{cases} \frac{4}{\pi} \int_0^{\pi/2} f(t) \sin(h\omega t) d(\omega t) & \text{for odd } h \\ 0 & \text{for even } h \end{cases}$ |

Even function $f(-t) = f(t)$

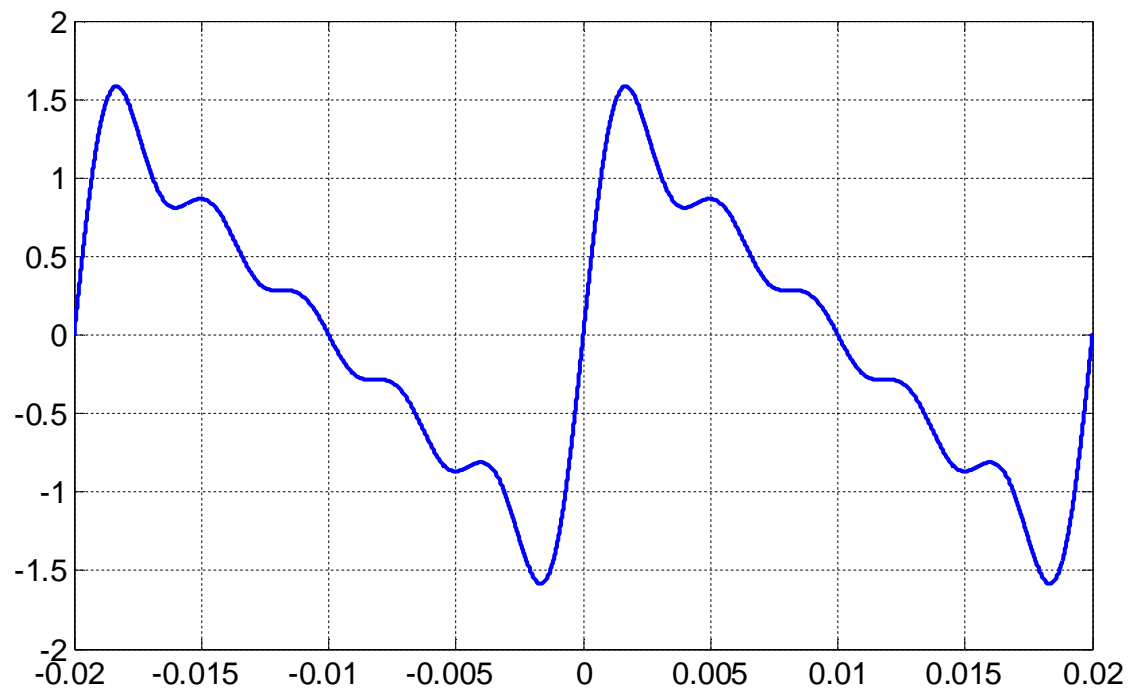
$$\cos(\omega t) + \frac{1}{2} \cos(2\omega t) + \frac{1}{3} \cos(3\omega t) + \frac{1}{4} \cos(4\omega t) + \frac{1}{5} \cos(5\omega t)$$



Odd function

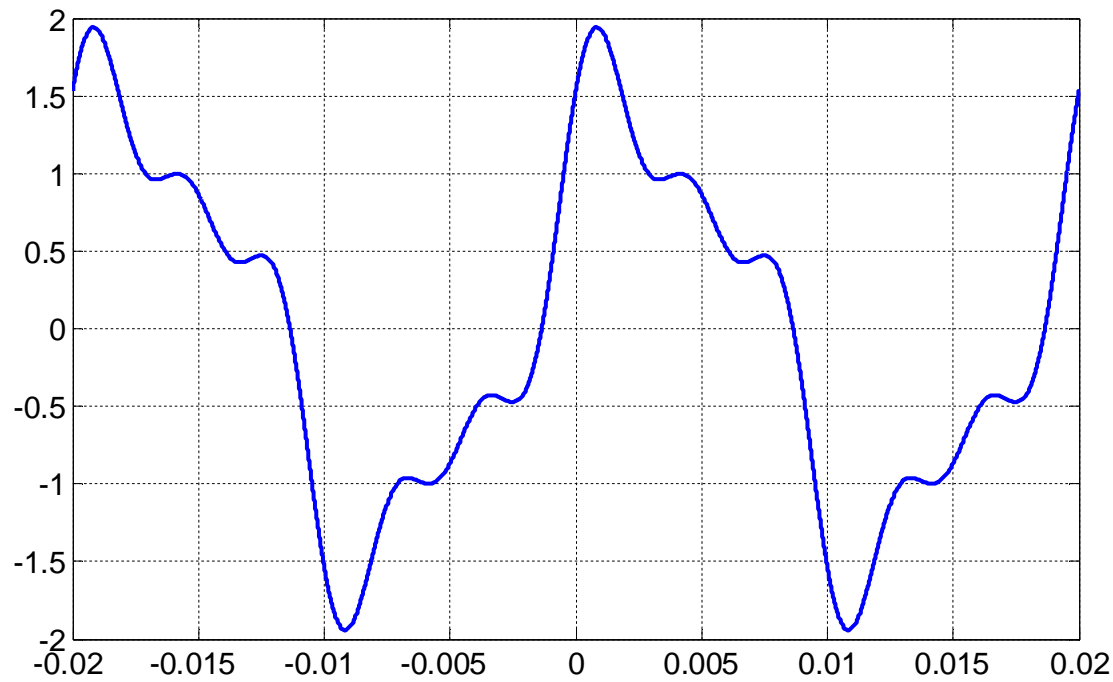
$$f(-t) = -f(t)$$

$$\sin(\omega t) + \frac{1}{2} \sin(2\omega t) + \frac{1}{3} \sin(3\omega t) + \frac{1}{4} \sin(4\omega t) + \frac{1}{5} \sin(5\omega t)$$



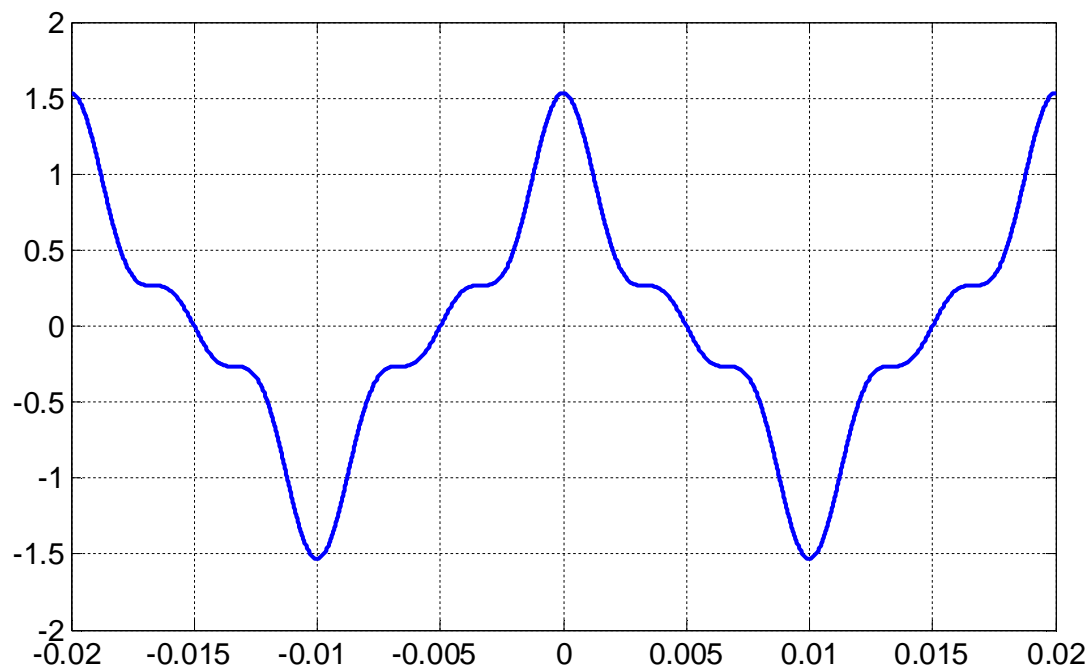
Half-wave symmetry $f(t) = -f(t + \frac{1}{2}T)$

$$\sin(\omega t) + \cos(\omega t) + \frac{1}{3}\sin(3\omega t) + \frac{1}{3}\cos(3\omega t) + \frac{1}{5}\sin(5\omega t) + \frac{1}{5}\cos(5\omega t)$$



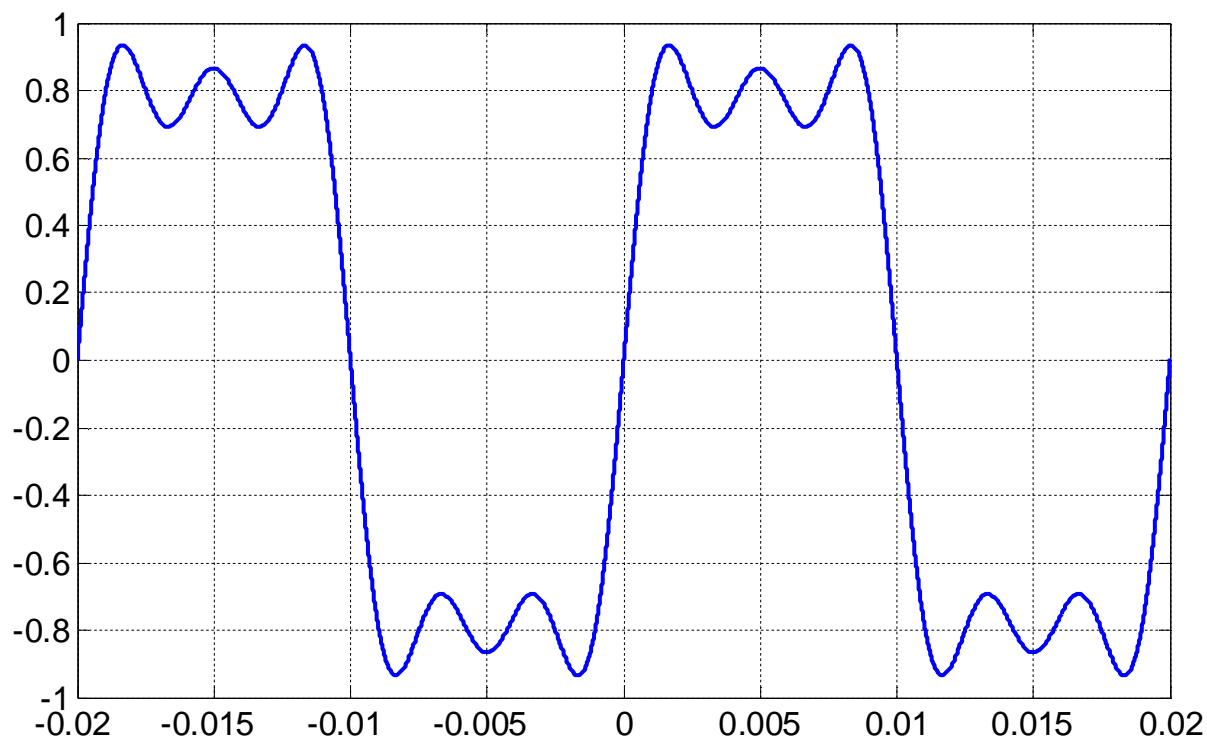
Even function, half-wave symmetry

$$\cos(\omega t) + \frac{1}{3}\cos(3\omega t) + \frac{1}{5}\cos(5\omega t)$$



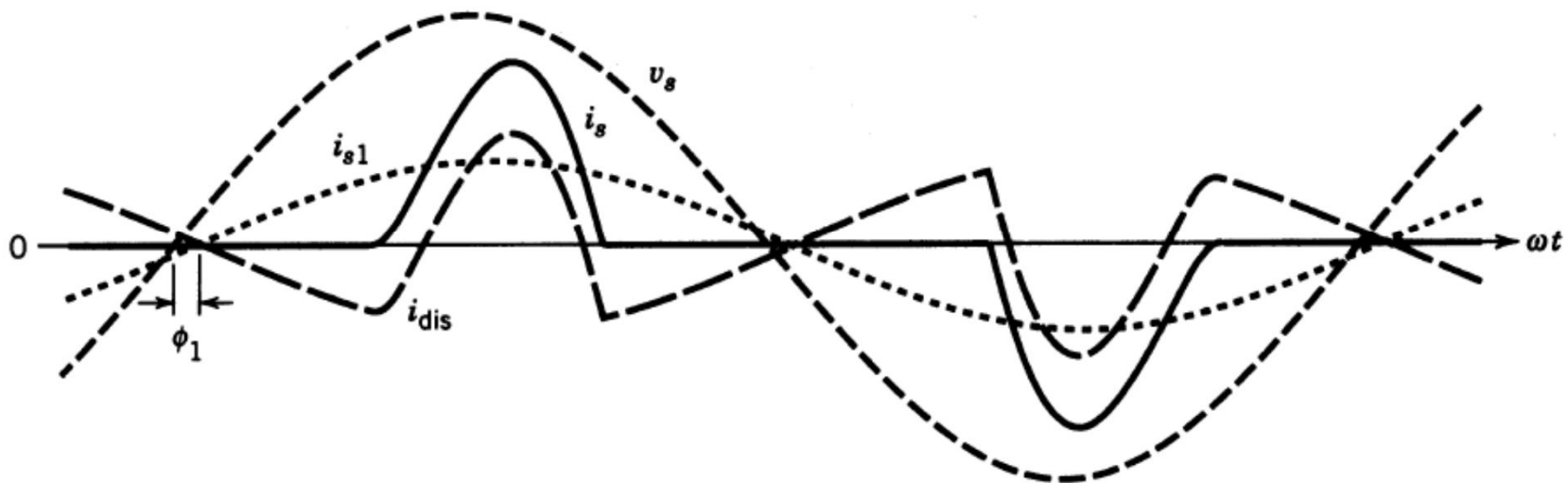
Odd function, half-wave symmetry

$$\sin(\omega t) + \frac{1}{3}\sin(3\omega t) + \frac{1}{5}\sin(5\omega t)$$



Total RMS incl harmonics

- $i_s(t) = i_{s1}(t) + \sum_{h=2}^n i_{sh}(t)$
- $I_s = \sqrt{\frac{1}{T_1} \int_0^{T_1} i_s^2(t) dt} = \sqrt{I_{s1}^2 + \sum_{h=2}^n I_{sh}^2} = \sqrt{I_{s1}^2 + I_{s3}^2 + I_{s5}^2 \dots}$
- (All cross-product terms, $i_{s1} \cdot i_{s2}, i_{s1} \cdot i_{s3} = 0$)



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