

EXAMINATION IN
-
TSEK03
RADIO FREQUENCY INTEGRATED CIRCUITS

Date: 2014-03-20
Time: 8-12
Location: U1
Aids: Calculator, Dictionary
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12 points are required to pass.

12-16 : 3

16-20 : 4

20-24 : 5

Please start each new problem at the top of a page!
Only use one side of each paper!

- 1) The mean square thermal noise density of a resistor in the room temperature is $33 \times 10^{-17} \text{ V}^2/\text{Hz}$. If this resistor is used in a first-order RC filter as shown in Fig. 1, and the noise bandwidth of the RC filter is 50 MHz , calculate the value of C in Fig.1. Present the details of your calculations. (4 p)

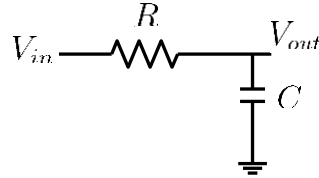


Fig. 1. A single-pole RC filter.

Hints:

- i) Boltzmann's constant is $1.38 \times 10^{-23} \text{ J/K}$
 ii) $\int \frac{dx}{1+x^2} = \tan^{-1} x$
 iii) Noise bandwidth: $\Delta f = \frac{1}{|H_{pk}|^2} \int_0^\infty |H(f)|^2 df$

- 2) Input stage of a single-ended LNA is shown in Fig. 2. Assume that $C_{GS1} = 1 \text{ pF}$, $L_1 = 1.6 \text{ nH}$, and $\lambda = \gamma = 0$.
- (a) Calculate the frequency at which the input impedance is purely resistive (a real value). (3 p)
- (b) Calculate the transconductance of the transistor (g_{m1}), to match with a $50\text{-}\Omega$ source resistance. (2 p)

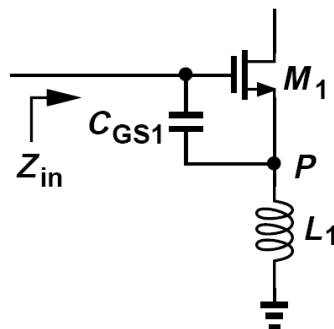


Fig. 2. Input stage of a single-ended LNA.

3) A single-balanced mixer is shown in Fig. 3. Ignore channel length modulation.

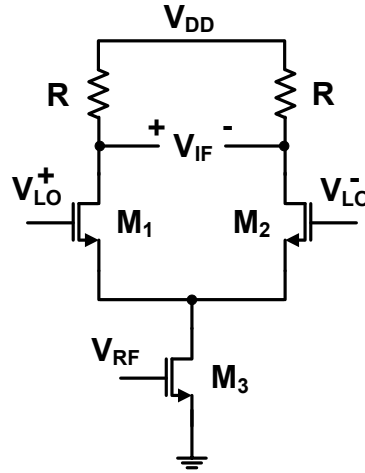


Fig. 3. A single-balanced mixer.

(a) If LO signal is a square wave toggling between 0 and 1 with 50% duty cycle and LO switching is abrupt, derive an expression for the conversion gain of this mixer. (3 p)

(b) If LO signal is a sine wave varying between 0 and 1, drain currents of M_1 and M_2 will remain approximately equal for a period of ΔT in each half cycle reducing the conversion gain. If ΔT is 10% of the LO signal period, calculate the conversion gain reduction in dB. (2 p)

Hint:

For a square wave LO signal toggling between -1 and 1:

$$V_{LO}(t) = \frac{4}{\pi} \cos \omega_{LO}(t) - \frac{4}{3\pi} \cos 3\omega_{LO}(t) + \frac{4}{5\pi} \cos 5\omega_{LO}(t) - \dots$$

4) Figure 4 shows a unity gain feedback system with two identical ideal integrators. The transfer function of each integrator in Laplace domain is $H(s) = K/s$.

(a) Is this feedback system stable? Motivate your answer using Barkhausen criteria. (2 p)

(b) Write a differential equation describing this feedback system and solve it for $X=0$. Is this result consistent with your answer in part (a)? Why? (3 p)

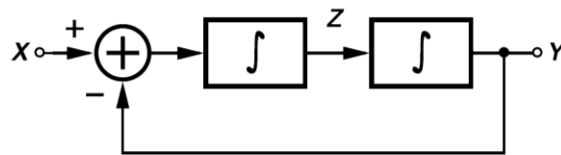


Fig. 4. Two ideal integrators inside a unity gain feedback.

5) Figure 5 shows a block level description of a PLL.

(a) Determine the closed-loop transfer function (i.e., $\frac{\Phi_{out}}{\Phi_{in}}(s)$) and the type of the PLL. (3 p)

(b) Prove that for slow input phase variations the output tracks the input. (2 p)

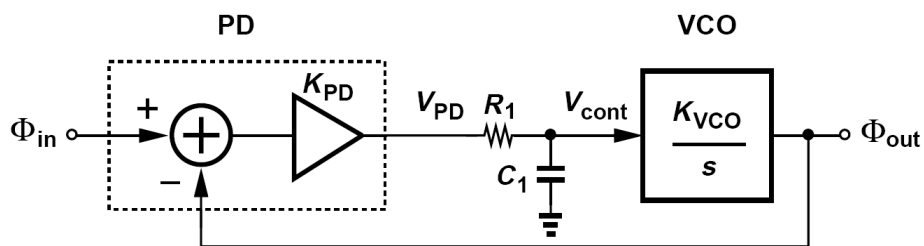
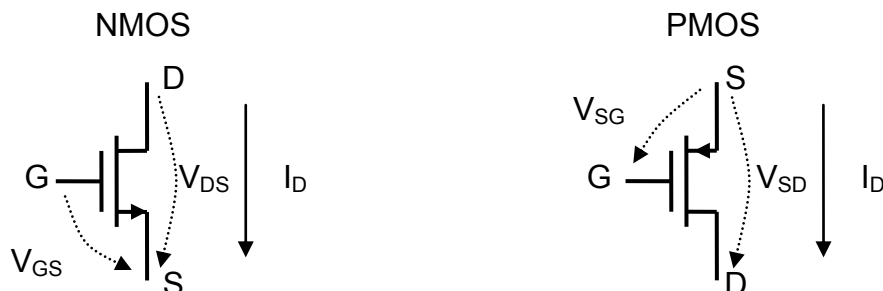


Fig. 5. Block diagram of a PLL.

TRANSISTOR EQUATIONS



NMOS

- **Cutoff:** $I_D = 0$ ($V_{GS} < V_{TN}$)

- **Linear mode:**

$$I_D = \mu_n C_{ox} \frac{W}{L} \left((V_{GS} - V_{TN}) V_{DS} - \frac{V_{DS}^2}{2} \right) \quad (V_{GS} > V_{TN}) \text{ and } (V_{DS} < V_{GS} - V_{TN})$$

- **Saturation mode:**

$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} (V_{GS} - V_{TN})^2 (1 + \lambda V_{DS}) \quad (V_{GS} > V_{TN}) \text{ and } (V_{DS} > V_{GS} - V_{TN})$$

PMOS

- **Cutoff:** $I_D = 0$ ($V_{GS} < |V_{TP}|$)

- **Linear mode:**

$$I_D = \mu_p C_{ox} \frac{W}{L} \left((V_{SG} - |V_{TP}|) V_{SD} - \frac{V_{SD}^2}{2} \right) \quad (V_{GS} > |V_{TP}|) \text{ and } (V_{SD} < V_{SG} - |V_{TP}|)$$

- **Saturation mode:**

$$I_D = \frac{1}{2} \mu_p C_{ox} \frac{W}{L} (V_{SG} - |V_{TP}|)^2 (1 + \lambda V_{SD}) \quad (V_{GS} > |V_{TP}|) \text{ and } (V_{SD} > V_{SG} - |V_{TP}|)$$